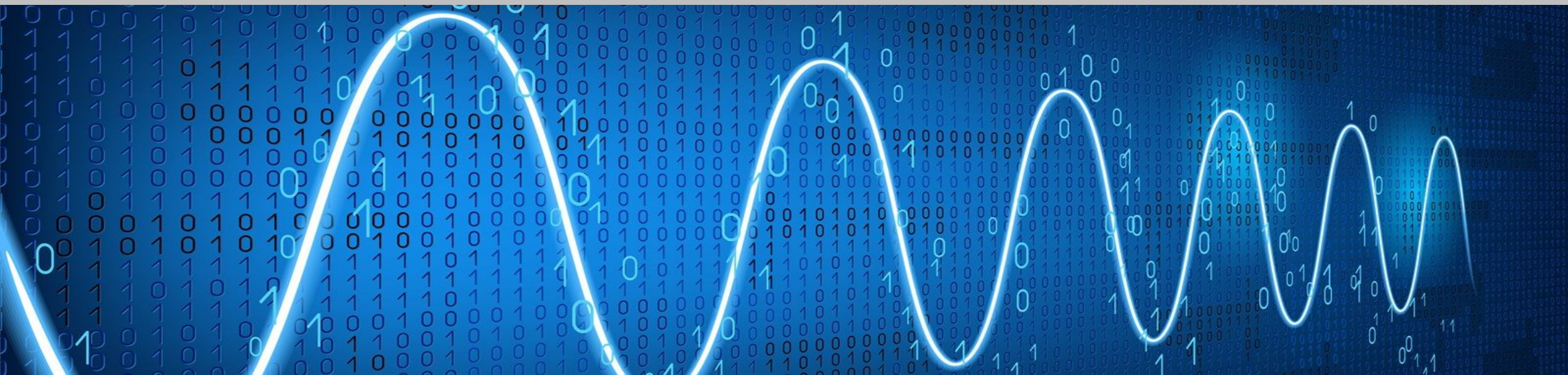


# Digital Signal Processing

Lab 06: Analyzing Discrete-Time Systems

Abdallah El Ghamry



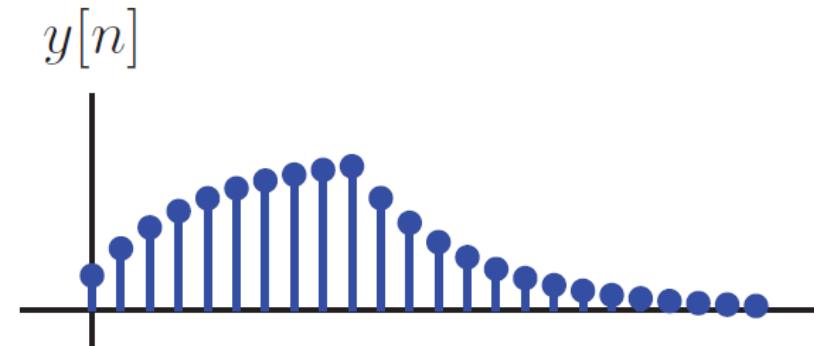
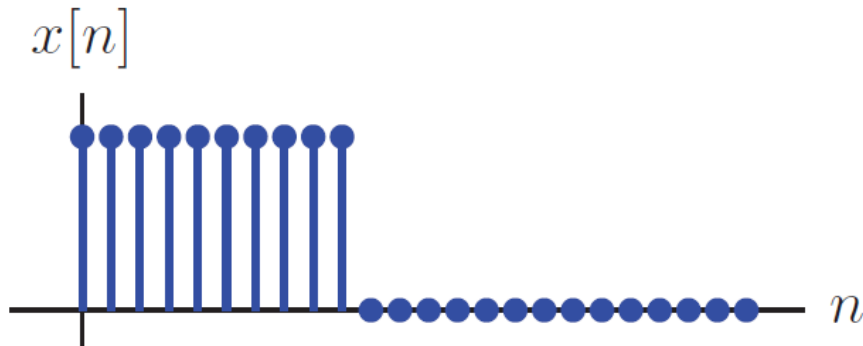
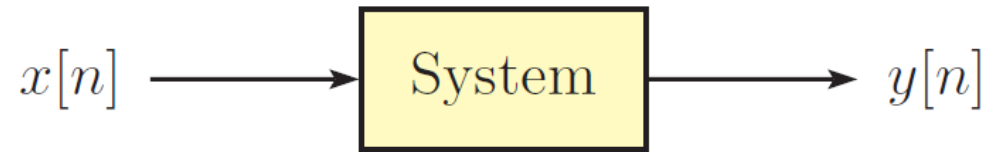
# Analyzing Discrete-Time Systems in the Time Domain

The purpose of this lab is to

- Develop the notion of a **discrete-time system**.
- Discuss the concepts of **linearity** and **time invariance**.
- Learn how to compute the output signal for a **linear and time-invariant system** using **convolution**.
- Understand the graphical interpretation of the steps involved in carrying out the **convolution operation**.
- Learn the concepts of **causality** and **stability**.

# Discrete-Time System

- A **discrete-time system** is a mathematical formula, method or algorithm that defines a **cause-effect relationship** between a set of discrete-time **input signals** and a set of discrete-time **output signals**.



# Discrete-Time System

- The **input-output relationship** of a discrete-time system may be expressed in the form

$$y[n] = \text{Sys}\{x[n]\}$$

- A system that simply **multiplies** its input signal by a **constant gain factor**  $K$

$$y[n] = Kx[n]$$

- A system that **delays** its input signal by  $m$  **samples**

$$y[n] = x[n - m]$$

- A system that produces an output signal **proportional to the square** of the input signal

$$y[n] = K[x[n]]^2$$

- Linearity property will be very important as we analyze and design discrete-time systems.
- Conditions for **linearity** of a discrete-time system are:

$$\text{Sys}\{x_1[n] + x_2[n]\} = \text{Sys}\{x_1[n]\} + \text{Sys}\{x_2[n]\}$$

$$\text{Sys}\{\alpha_1 x_1[n]\} = \alpha_1 \text{Sys}\{x_1[n]\}$$

- The **additivity rule** can be stated as follows:

The response of a linear system to **the sum of two signals** is **the same as the sum of individual responses** to each of the two input signals.

$$\text{Sys}\{x_1[n] + x_2[n]\} = \text{Sys}\{x_1[n]\} + \text{Sys}\{x_2[n]\}$$

- The **homogeneity rule** can be stated as follows:

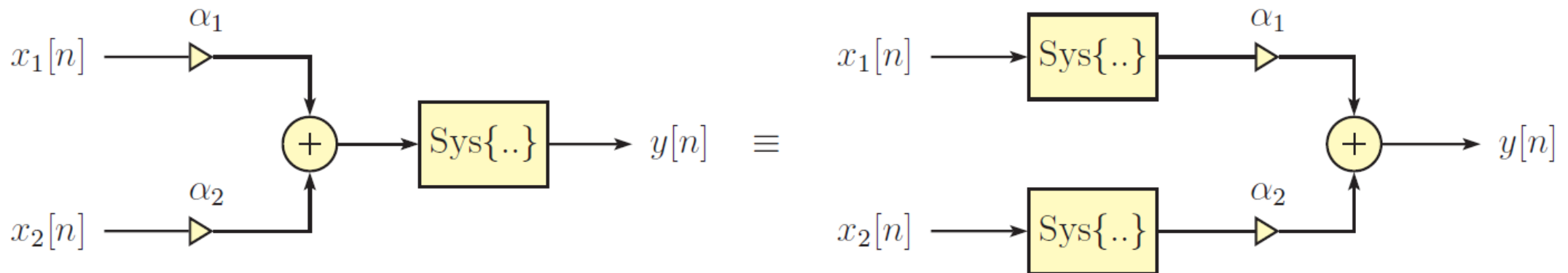
**Scaling the input signal** of a linear system by a constant **gain factor** causes the output signal to be **scaled with the same gain factor**.

$$\text{Sys}\{\alpha_1 x_1[n]\} = \alpha_1 \text{Sys}\{x_1[n]\}$$

# Linearity: Superposition Principle

- The two criteria can be combined into one equation which is referred to as the **superposition principle**.
- The response of the system to a **weighted sum of any two input signals** is **equal to** the **same weighted sum of the individual responses** of the system to each of the two input signals.

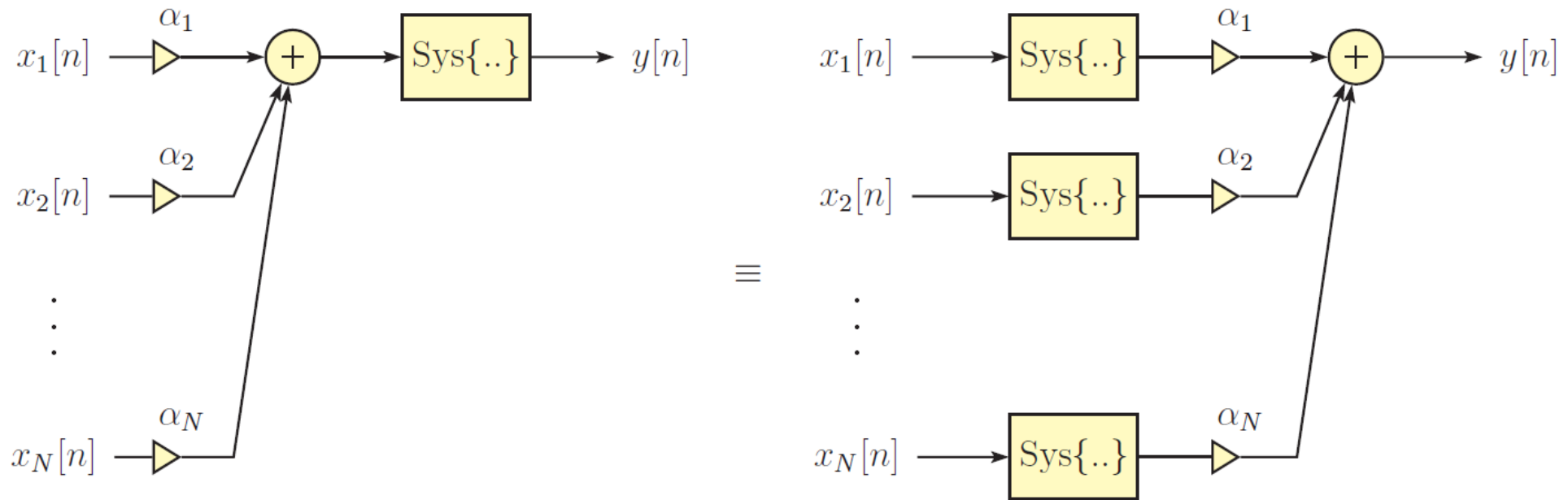
$$\text{Sys}\{\alpha_1 x_1[n] + \alpha_2 x_2[n]\} = \alpha_1 \text{Sys}\{x_1[n]\} + \alpha_2 \text{Sys}\{x_2[n]\}$$



# Linearity: Superposition Principle

- A **generalization** of the principle of superposition for the weighted sum of  $N$  discrete-time signals is expressed as

$$y[n] = \text{Sys} \left\{ \sum_{i=1}^N \alpha_i x_i[n] \right\} = \sum_{i=1}^N \alpha_i y_i[n]$$





## Example 3.1

For each of the discrete-time systems described below, determine whether the system is **linear or not**:

a.  $y[n] = 3x[n] + 2x[n - 1]$

b.  $y[n] = 3x[n] + 2x[n + 1]x[n - 1]$

c.  $y[n] = a^{-n}x[n]$

## Example 3.1 (a) – Solution

- a. In order to test the linearity of the system we will think of its responses to the two discrete-time signals  $x_1[n]$  and  $x_2[n]$  as

$$y_1[n] = \text{Sys} \{x_1[n]\} = 3x_1[n] + 2x_1[n-1]$$

and

$$y_2[n] = \text{Sys} \{x_2[n]\} = 3x_2[n] + 2x_2[n-1]$$

The response of the system to the linear combination signal  $x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$  is computed as

$$\begin{aligned} y[n] &= \text{Sys} \{ \alpha_1 x_1[n] + \alpha_2 x_2[n] \} \\ &= 3 (\alpha_1 x_1[n] + \alpha_2 x_2[n]) + 2 (\alpha_1 x_1[n-1] + \alpha_2 x_2[n-1]) \\ &= \alpha_1 (3x_1[n] + 2x_1[n-1]) + \alpha_2 (3x_2[n] + 2x_2[n-1]) \\ &= \alpha_1 y_1[n] + \alpha_2 y_2[n] \end{aligned}$$

Superposition principle holds, and therefore the system in question is linear.

## Example 3.1 (b) – Solution

b. Again using the test signals  $x_1[n]$  and  $x_2[n]$  we have

$$y_1[n] = \text{Sys} \{x_1[n]\} = 3x_1[n] + 2x_1[n+1]x_1[n-1]$$

and

$$y_2[n] = \text{Sys} \{x_2[n]\} = 3x_2[n] + 2x_2[n+1]x_2[n-1]$$

Use of the linear combination signal  $x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$  as input to the system yields the output signal

$$\begin{aligned} y[n] &= \text{Sys} \{ \alpha_1 x_1[n] + \alpha_2 x_2[n] \} \\ &= 3 (\alpha_1 x_1[n] + \alpha_2 x_2[n]) \\ &\quad + 2 (\alpha_1 x_1[n+1] + \alpha_2 x_2[n+1]) (\alpha_1 x_1[n-1] + \alpha_2 x_2[n-1]) \end{aligned}$$

In this case the superposition principle does not hold true. The system in part (b) is therefore not linear.

## Example 3.1 (c) – Solution

c. The responses of the system to the two test signals are

$$y_1[n] = \text{Sys} \{x_1[n]\} = a^{-n} x_1[n]$$

and

$$y_2[n] = \text{Sys} \{x_2[n]\} = a^{-n} x_2[n]$$

and the response to the linear combination signal  $x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$  is

$$\begin{aligned} y[n] &= \text{Sys} \{ \alpha_1 x_1[n] + \alpha_2 x_2[n] \} \\ &= a^{-n} (\alpha_1 x_1[n] + \alpha_2 x_2[n]) \\ &= \alpha_1 a^{-n} x_1[n] + \alpha_2 a^{-n} x_2[n] \\ &= \alpha_1 y_1[n] + \alpha_2 y_2[n] \end{aligned}$$

The system is linear.

# Time Invariance

- Let a discrete-time system be described with the input-output relationship

$$y[n] = \text{Sys}\{x[n]\}$$

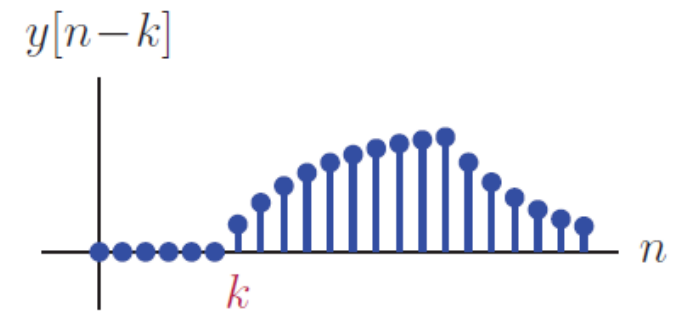
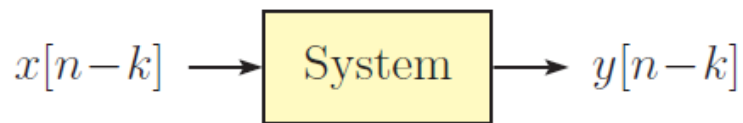
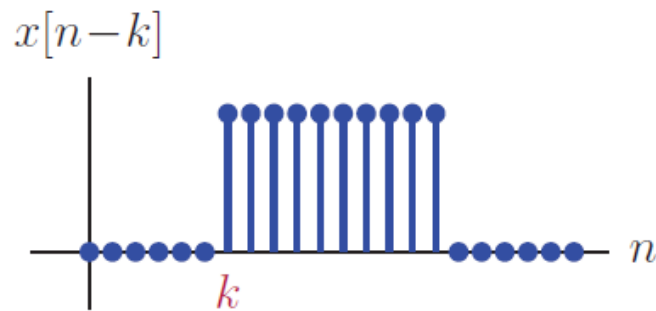
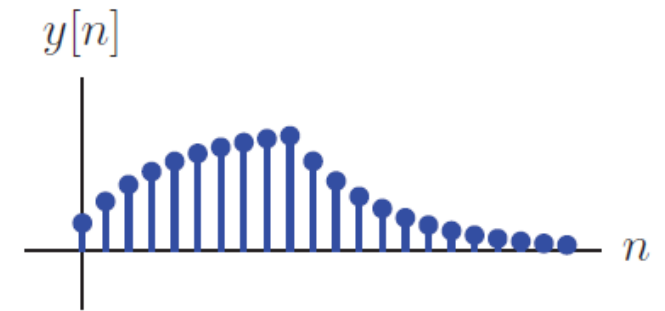
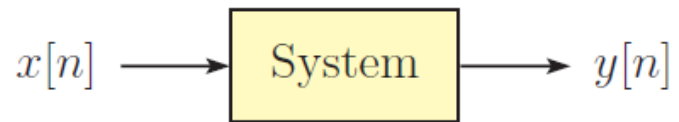
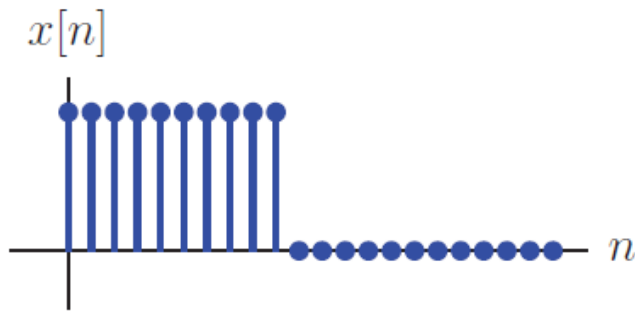
- For the system to be considered **time-invariant**, the only effect of time-shifting the input signal should be to cause an equal amount of time shift in the output signal.

$$\text{Sys}\{x[n]\} = y[n] \quad \text{implies that} \quad \text{Sys}\{x[n - k]\} = y[n - k]$$

# Time Invariance

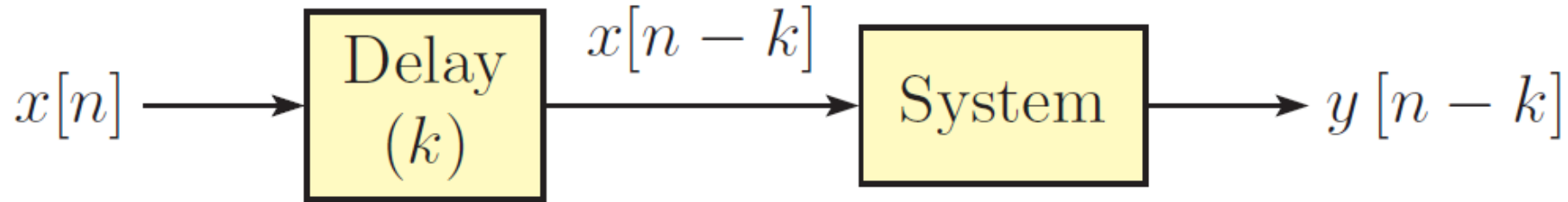
- Condition for **time-invariance**:

$$\text{Sys} \{x[n]\} = y[n] \quad \text{implies that} \quad \text{Sys} \{x[n - k]\} = y[n - k]$$

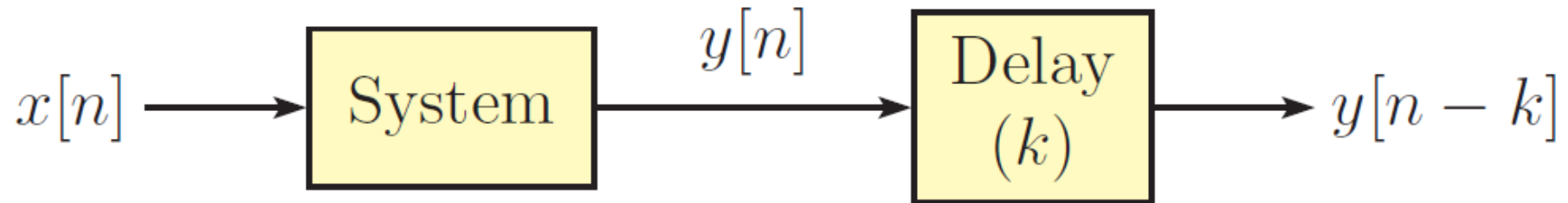


# Time Invariance

- The **time-invariant nature** of a system can be characterized by the equivalence of the two configurations shown in Figure.



(a)



## Example 3.2

For each of the discrete-time systems described below, determine whether the system **is time-invariant or not**:

a.  $y[n] = y[n - 1] + 3x[n]$

b.  $y[n] = x[n]y[n - 1]$

c.  $y[n] = nx[n - 1]$



## Example 3.2 (a) – Solution

- a. We will test the time-invariance property of the system by time-shifting both the input and the output signals by the same number of samples, and see if the input–output relationship still holds. Replacing the index  $n$  by  $n - k$  in the arguments of all input and output terms we obtain

$$\text{Sys} \{x[n - k]\} = y[n - k - 1] + 3x[n - k] = y[n - k]$$

The input–output relationship holds, therefore the system is time-invariant.

## Example 3.2 (b) – Solution

b. Proceeding in a similar fashion we have

$$\text{Sys} \{x[n - k]\} = x[n - k] y[n - k - 1] = y[n - k]$$

This system is time-invariant as well.

## Example 3.2 (c) – Solution

- c. Replacing the index  $n$  by  $n - k$  in the arguments of all input and output terms yields

$$\text{Sys} \{x[n - k]\} = n x[n - k - 1] \neq y[n - k]$$

This system is clearly not time-invariant since the input–output relationship no longer holds after input and output signals are time-shifted.

Should we have included the factor  $n$  in the time shifting operation when we wrote the response of the system to a time-shifted input signal? In other words, should we have written the response as

$$\text{Sys} \{x[n - k]\} \stackrel{?}{=} (n - k) x[n - k - 1]$$

The answer is no. The factor  $n$  that multiplies the input signal is part of the system definition and not part of either the input or the output signal. Therefore we cannot include it in the process of time-shifting input and output signals.

## Problem 3.1

A number of discrete-time systems are specified below in terms of their input-output relationships.

For each case determine if the system is **linear and/or time-invariant**.

**a.**  $y[n] = x[n] u[n]$

**c.**  $y[n] = 3x[n] + 5u[n]$

**e.**  $y[n] = \cos(0.2\pi n) x[n]$

**f.**  $y[n] = x[n] + 3x[n - 1]$

## Problem 3.1 (a) – Solution

**a.**  $y[n] = x[n] u[n]$

**a.**

$$y_1[n] = \text{Sys} \{x_1[n]\} = x_1[n] u[n]$$

$$y_2[n] = \text{Sys} \{x_2[n]\} = x_2[n] u[n]$$

Using  $x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$  as input we obtain

$$\begin{aligned} y[n] &= \text{Sys} \{ \alpha_1 x_1[n] + \alpha_2 x_2[n] \} \\ &= (\alpha_1 x_1[n] + \alpha_2 x_2[n]) u[n] \\ &= \alpha_1 y_1[n] + \alpha_2 y_2[n] \end{aligned}$$

The system is linear.

$$\text{Sys} \{x_1[n - m]\} = x_1[n - m] u[n] \neq y_1[n - m]$$

The system is not time-invariant.

## Problem 3.1 (c) – Solution

**c.**  $y[n] = 3x[n] + 5u[n]$

**c.**

$$y_1[n] = \text{Sys}\{x_1[n]\} = 3x_1[n] + 5u[n]$$

$$y_2[n] = \text{Sys}\{x_2[n]\} = 3x_2[n] + 5u[n]$$

Using  $x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$  as input we obtain

$$\begin{aligned} y[n] &= \text{Sys}\{\alpha_1 x_1[n] + \alpha_2 x_2[n]\} \\ &= 3(\alpha_1 x_1[n] + \alpha_2 x_2[n]) + 5u[n] \\ &\neq \alpha_1 y_1[n] + \alpha_2 y_2[n] \end{aligned}$$

The system is not linear.

$$\text{Sys}\{x_1[n-m]\} = 3x_1[n-m] + 5u[n] \neq y_1[n-m]$$

The system is not time-invariant.

## Problem 3.1 (e) – Solution

e.  $y[n] = \cos(0.2\pi n) x[n]$

e.

$$y_1[n] = \text{Sys} \{x_1[n]\} = \cos(0.2\pi n) x_1[n]$$

$$y_2[n] = \text{Sys} \{x_2[n]\} = \cos(0.2\pi n) x_2[n]$$

Using  $x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$  as input we obtain

$$\begin{aligned} y[n] &= \text{Sys} \{ \alpha_1 x_1[n] + \alpha_2 x_2[n] \} \\ &= \cos(0.2\pi n) (\alpha_1 x_1[n] + \alpha_2 x_2[n]) \\ &= \alpha_1 y_1[n] + \alpha_2 y_2[n] \end{aligned}$$

The system is linear.

$$\text{Sys} \{x_1[n - m]\} = \cos(0.2\pi n) x_1[n - m] \neq y_1[n - m]$$

The system is not time-invariant.

## Problem 3.1 (f) – Solution

$$\mathbf{f.} \quad y[n] = x[n] + 3x[n-1]$$

**f.**

$$y_1[n] = \text{Sys} \{x_1[n]\} = x_1[n] + 3x_1[n-1]$$

$$y_2[n] = \text{Sys} \{x_2[n]\} = x_2[n] + 3x_2[n-1]$$

Using  $x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$  as input we obtain

$$\begin{aligned} y[n] &= \text{Sys} \{ \alpha_1 x_1[n] + \alpha_2 x_2[n] \} \\ &= (\alpha_1 x_1[n] + \alpha_2 x_2[n]) + 3(\alpha_1 x_1[n-1] + \alpha_2 x_2[n-1]) \\ &= \alpha_1 y_1[n] + \alpha_2 y_2[n] \end{aligned}$$

The system is linear.

$$\text{Sys} \{x_1[n-m]\} = x_1[n-m] + 3x_1[n-m-1] = y_1[n-m]$$

The system is time-invariant.



## Problem 3.2 (b)

Determine if the system is **linear or not**.

**b.** 
$$y[n] = \sum_{k=0}^n x[k]$$

## Problem 3.2 (b) – Solution

**b.**

$$y_1[n] = \text{Sys} \{x_1[n]\} = \sum_{k=0}^n x_1[k]$$

$$y_2[n] = \text{Sys} \{x_2[n]\} = \sum_{k=0}^n x_2[k]$$

Using  $x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$  as input we obtain

$$\begin{aligned} y[n] &= \text{Sys} \{ \alpha_1 x_1[n] + \alpha_2 x_2[n] \} \\ &= \sum_{k=0}^n (\alpha_1 x_1[k] + \alpha_2 x_2[k]) \\ &= \alpha_1 \sum_{k=0}^n x_1[k] + \alpha_2 \sum_{k=0}^n x_2[k] \\ &= \alpha_1 y_1[n] + \alpha_2 y_2[n] \end{aligned}$$

The system is linear.

## Problem 3.3

**3.3.** Consider the cascade combination of two systems shown in Fig. P.3.3(a).



**Figure P. 3.3**

**a.** Let the input-output relationships of the two subsystems be given as

$$\text{Sys}_1 \{x[n]\} = 3x[n] \quad \text{and} \quad \text{Sys}_2 \{w[n]\} = w[n - 2]$$

Write the relationship between  $x[n]$  and  $y[n]$ .

**b.** Let the order of the two subsystems be changed as shown in Fig. P.3.3(b). Write the relationship between  $x[n]$  and  $\bar{y}[n]$ . Does changing the order of two subsystems change the overall input-output relationship of the system?

## Problem 3.3 – Solution

**a.**

$$w[n] = 3x[n]$$

$$y[n] = w[n - 2] = 3x[n - 2]$$

**b.**

$$\bar{w}[n] = x[n - 2]$$

$$\bar{y}[n] = 3\bar{w}[n] = 3x[n - 2]$$

Input-output relationship of the system **does not change** when the order of the two subsystems is changed.

- Discrete-time systems that are **both linear and time-invariant** will play an important role in the rest of this textbook.
- We will develop **time- and frequency-domain analysis and design** techniques for working with such systems.
- To simplify the terminology, we will use the acronym **DTLTI** to refer to **discrete-time linear and time-invariant systems**.

# Difference Equations for Discrete-Time Systems

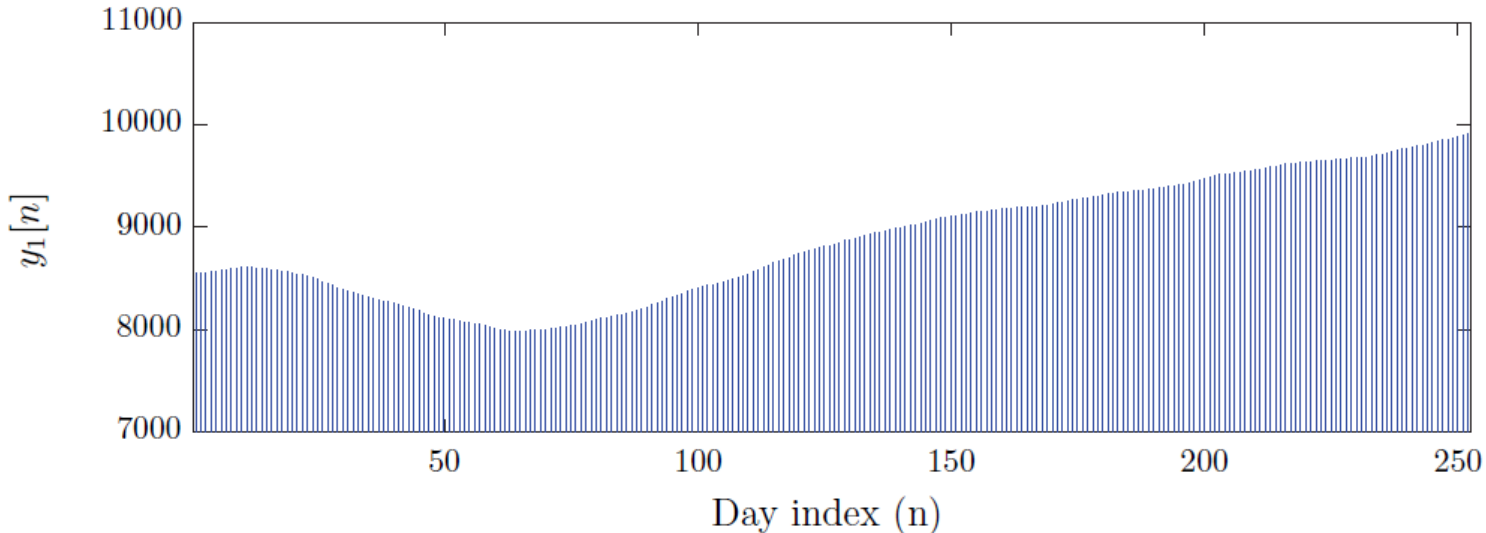
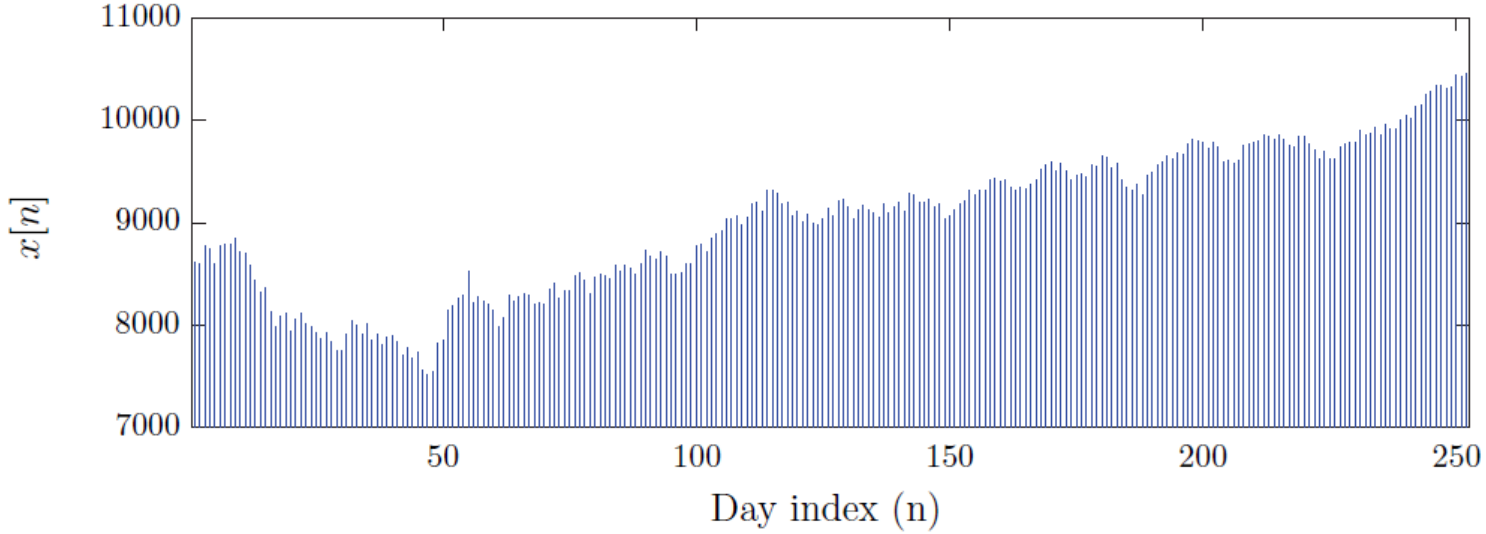
- In chapter 2, we have discussed methods of representing **continuous-time systems with differential equations**.
- Using a **similar approach**, discrete-time systems can be modeled with **difference equations** involving current, past, or future samples of input and output signals.
- We will focus on **difference equations for DTLTI systems**.

# Moving-Average Filter

- A **length- $N$  moving average filter** is a simple system that produces an output equal to the arithmetic average of the most recent  $N$  samples of the input signal.
- The general expression for the **length- $N$  moving average filter** is

$$y[n] = \frac{x[n] + x[n-1] + \dots + x[n-(N-1)]}{N}$$
$$= \frac{1}{N} \sum_{k=0}^{N-1} x[n-k]$$

# Moving-Average Filter





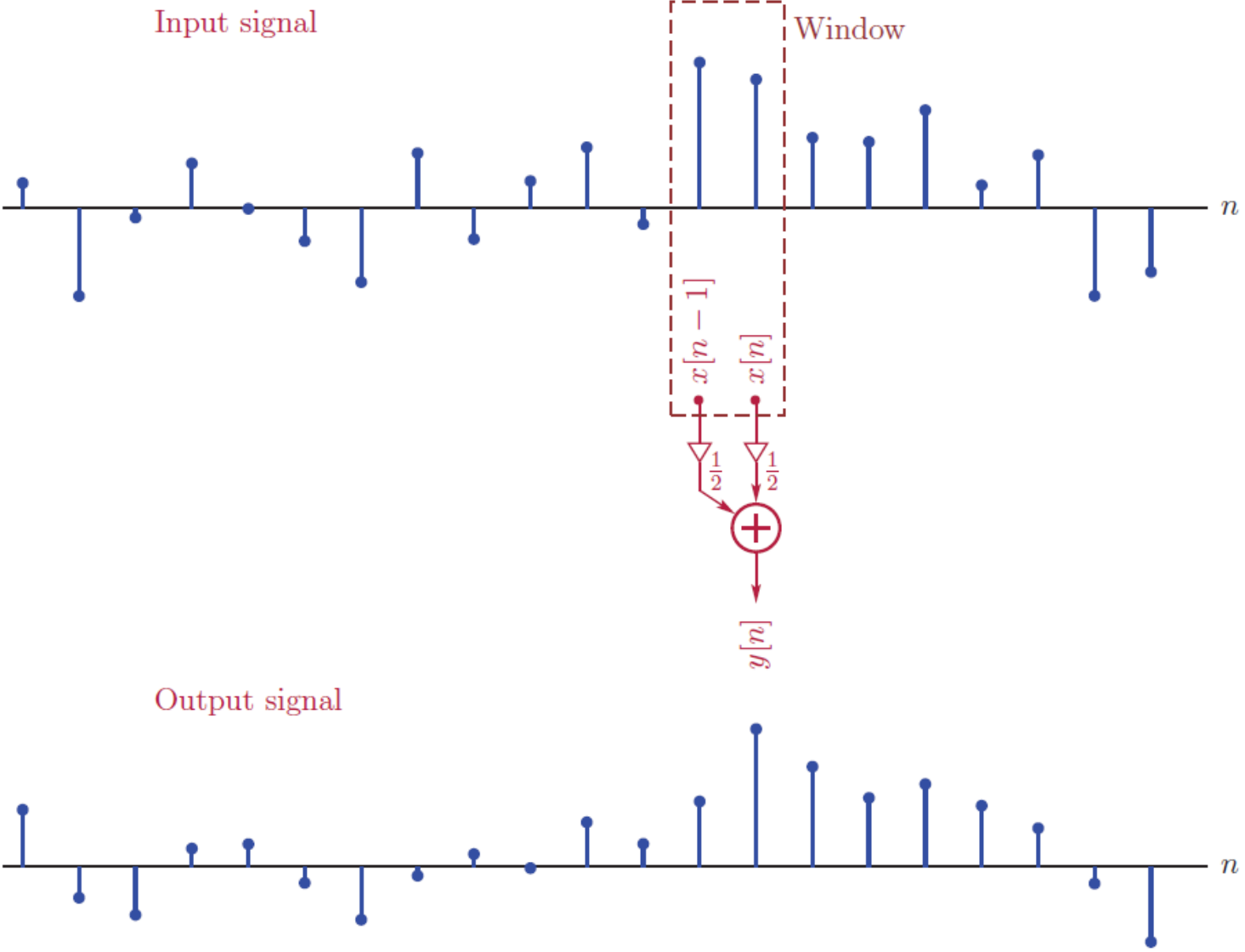
## Example 3.4: Length-2 Moving-Average Filter

- A **length-2 moving average** filter produces an output by **averaging** the **current input sample** and the **previous input sample**.
- This action translates to a difference equation in the form

$$y[n] = \frac{x[n] + x[n-1]}{2}$$

$$y[n] = \frac{1}{2}x[n] + \frac{1}{2}x[n-1]$$

# Example 3.4: Length-2 Moving-Average Filter



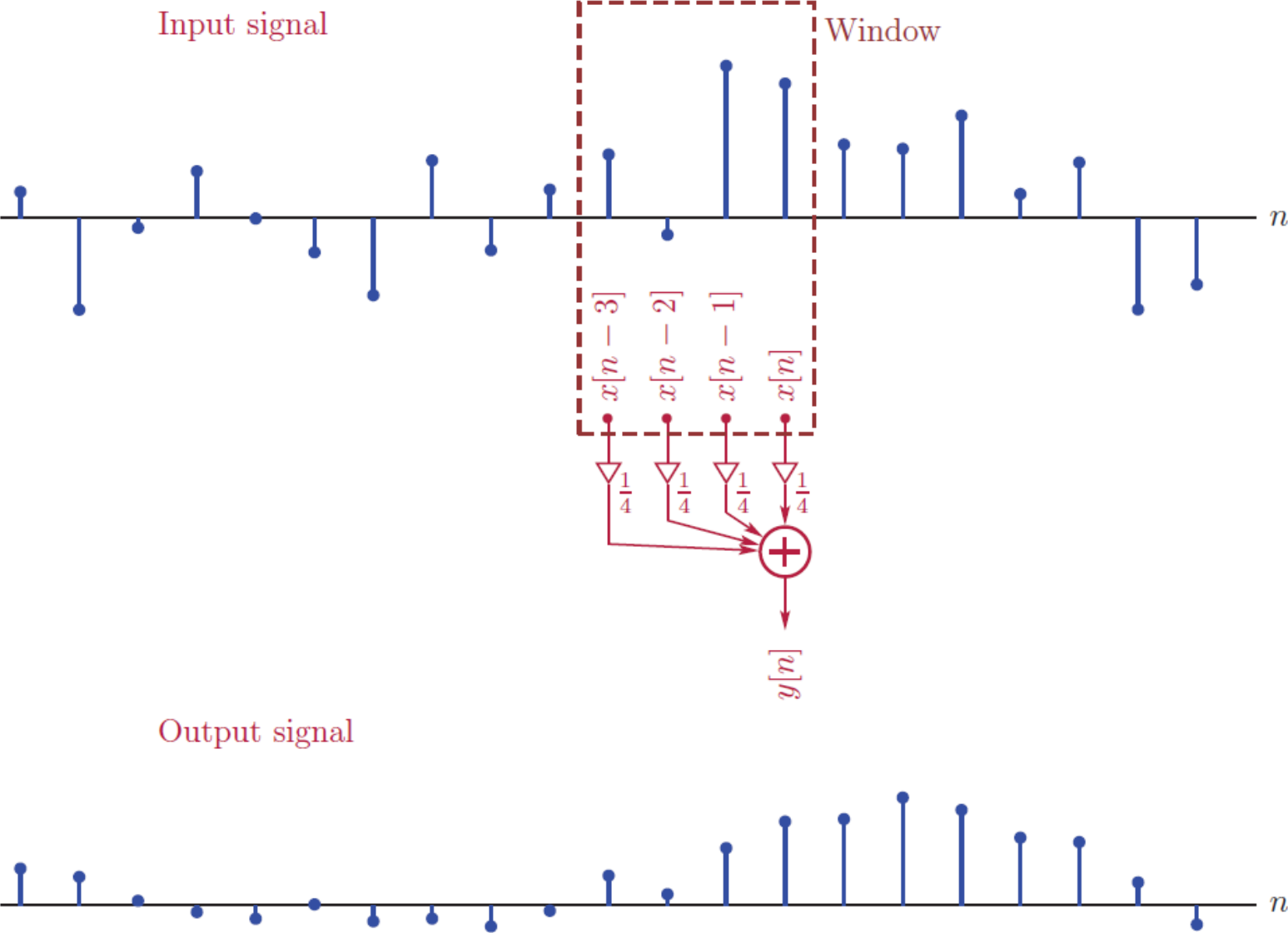
## Example 3.4: Length-4 Moving-Average Filter

- A **length-4 moving average filter** produces an output by **averaging** the **current input sample** and the **previous three input samples**.
- This action translates to a difference equation in the form

$$y[n] = \frac{x[n] + x[n-1] + x[n-2] + x[n-3]}{4}$$

$$y[n] = \frac{1}{4}x[n] + \frac{1}{4}x[n-1] + \frac{1}{4}x[n-2] + \frac{1}{4}x[n-3]$$

# Example 3.4: Length-4 Moving-Average Filter



## Problem 3.7

**3.7.** The discrete-time signal

$$x[n] = \{ \underset{\substack{\uparrow \\ n=0}}{1.7}, 2.3, 3.1, 3.3, 3.7, 2.9, 2.2, 1.4, 0.6, -0.2, 0.4 \}$$

is used as input to a length-2 moving average filter. Determine the response  $y[n]$  for  $n = 0, \dots, 9$ . Use  $x[-1] = 0$ .

## Problem 3.7 – Solution

$$y[0] = \frac{x[0] + x[-1]}{2} = \frac{1.7 + 0}{2} = 0.85$$

$$y[1] = \frac{x[1] + x[0]}{2} = \frac{2.3 + 1.7}{2} = 2$$

$$y[2] = \frac{x[2] + x[1]}{2} = \frac{3.1 + 2.3}{2} = 2.7$$

$$y[3] = \frac{x[3] + x[2]}{2} = \frac{3.3 + 3.1}{2} = 3.2$$

$$y[4] = \frac{x[4] + x[3]}{2} = \frac{3.7 + 3.3}{2} = 3.5$$

$$y[5] = \frac{x[5] + x[4]}{2} = \frac{2.9 + 3.7}{2} = 3.3$$

$$y[6] = \frac{x[6] + x[5]}{2} = \frac{2.2 + 2.9}{2} = 2.55$$

$$y[7] = \frac{x[7] + x[6]}{2} = \frac{1.4 + 2.2}{2} = 1.8$$

$$y[8] = \frac{x[8] + x[7]}{2} = \frac{0.6 + 1.4}{2} = 1$$

$$y[9] = \frac{x[9] + x[8]}{2} = \frac{-0.2 + 0.6}{2} = 0.2$$

$$y[10] = \frac{x[10] + x[9]}{2} = \frac{0.4 - 0.2}{2} = 0.1$$

$$y[n] = \{0.85, 2, 2.7, 3.2, 3.5, 3.3, 2.55, 1.8, 1, 0.2, 0.1\}$$

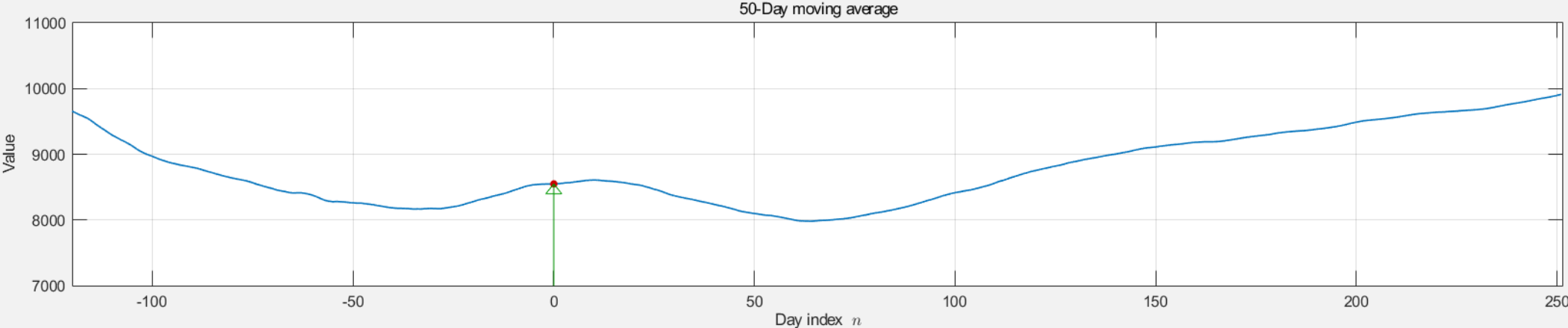
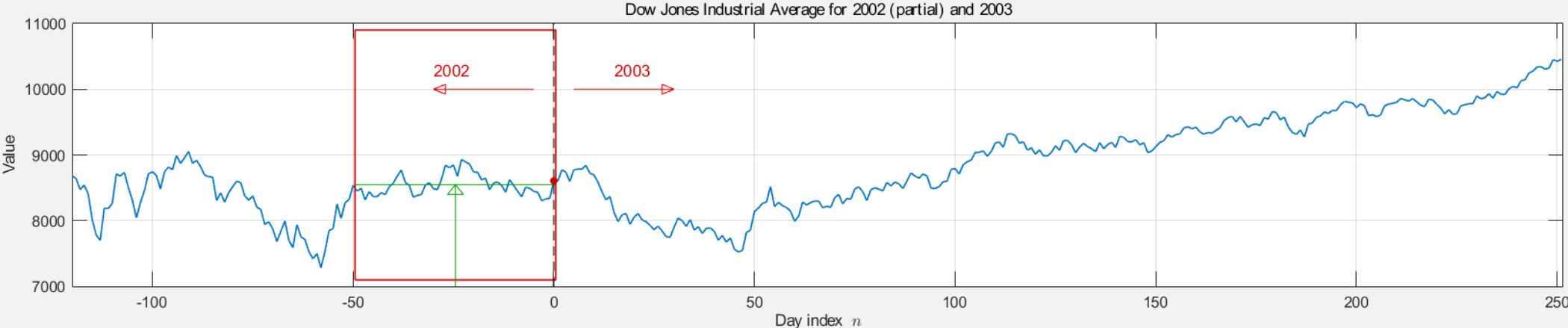
$\uparrow$   
 $n=0$

# Interactive Demo: ma\_demo1

Current index n:

Filter length (N):

Refer to: Section 3.3, Pages 191 through 194, Example 3.3, Eqn. (3.11), Figs. 3.6 and 3.7.



# Interactive Demo: ma\_demo2

## Length-2 Moving Average Filter

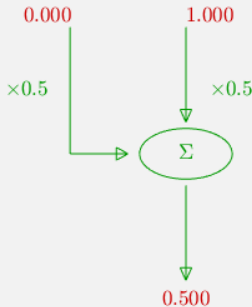
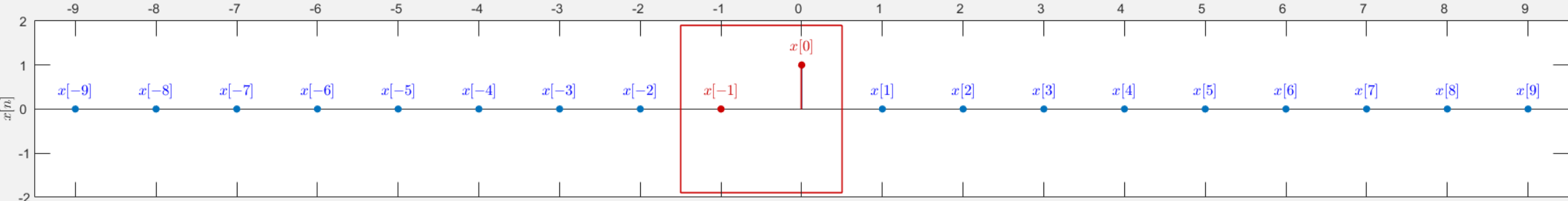
Refer to: Section 3.3, Pages 191 through 195, Example 3.4, Eqn. (3.12), Fig. 3.8.

Current index n:

< 0 >

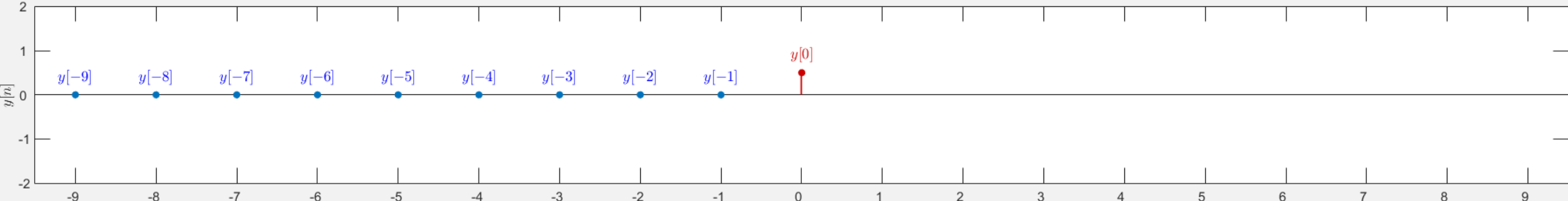
Input signal:

Unit-impulse signal



$$y[0] = 0.5x[0] + 0.5x[-1]$$

$x[-1] = 0.000$   
 $x[0] = 1.000$   
 $y[0] = 0.500$





# Interactive Demo: ma\_demo3

## Length-4 Moving Average Filter

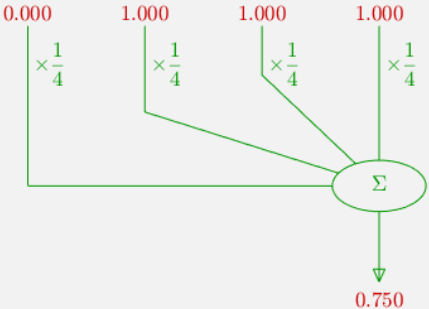
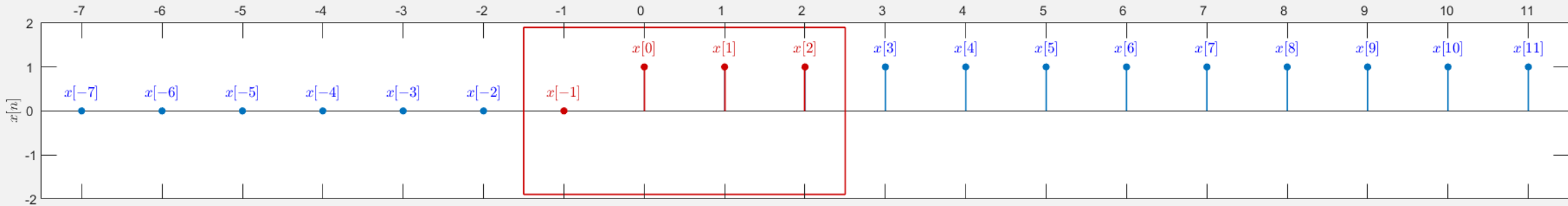
Refer to: Section 3.3, Pages 191 through 196, Example 3.5, Eqn. (3.13), Fig. 3.9.

Current index n:

<  >

Input signal:

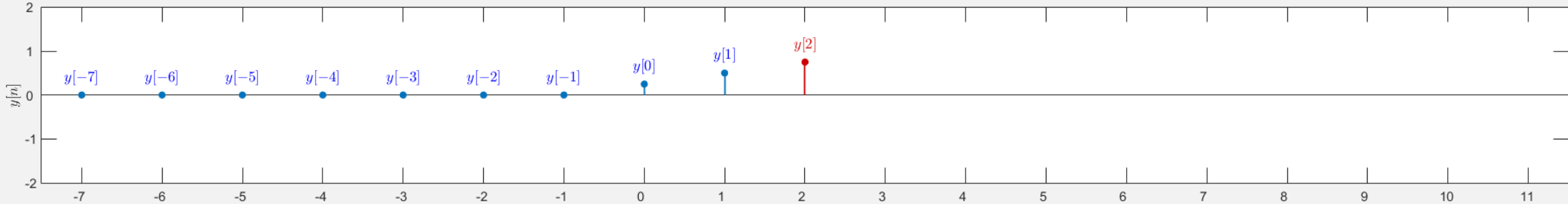
Unit-step signal



$$y[2] = \frac{1}{4}x[2] + \frac{1}{4}x[1] + \frac{1}{4}x[0] + \frac{1}{4}x[-1]$$

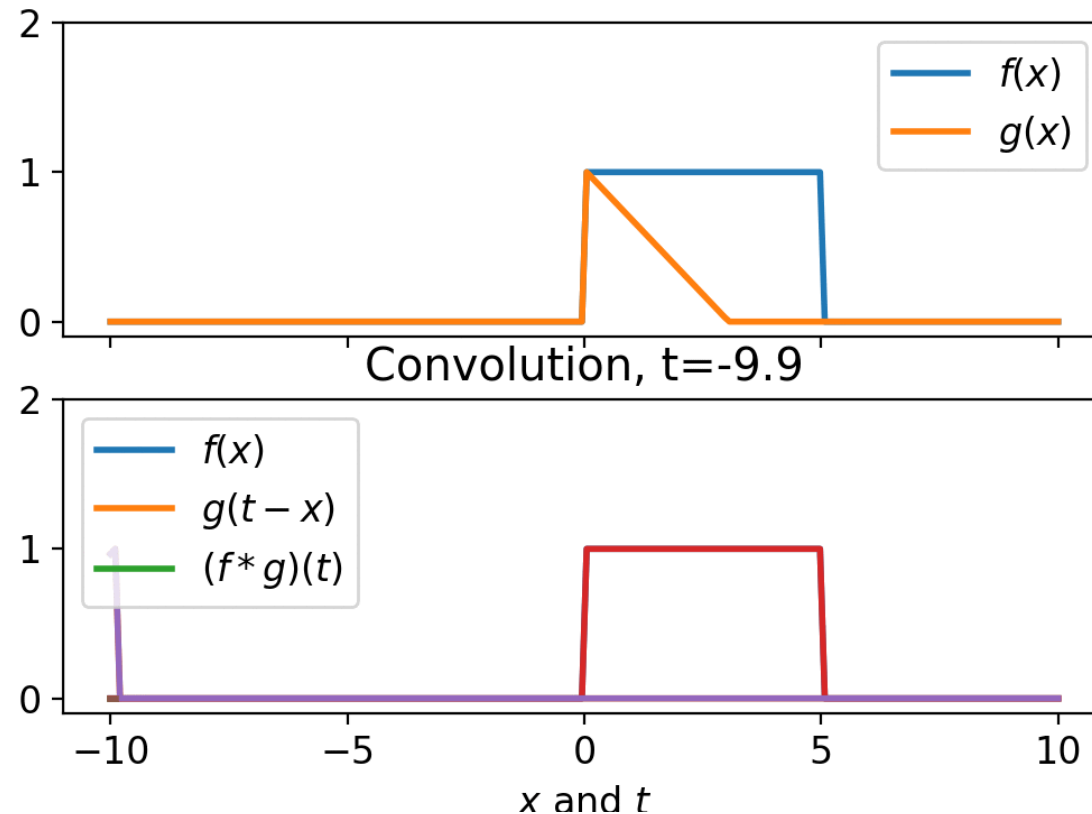
- $x[-1] = 0.000$
- $x[0] = 1.000$
- $x[1] = 1.000$
- $x[2] = 1.000$

$$y[2] = 0.750$$



# Continuous-Time Convolution

- A **convolution** is an **integral** that expresses the amount of overlap of one function when it is **shifted** over another function.



# Image Processing: Convolution vs. Correlation

- **Correlation** consists of moving the center of a kernel over an image, and computing the sum of products at each location.
- **Convolution** is the same as correlation, except that the correlation **kernel is rotated by 180°**.

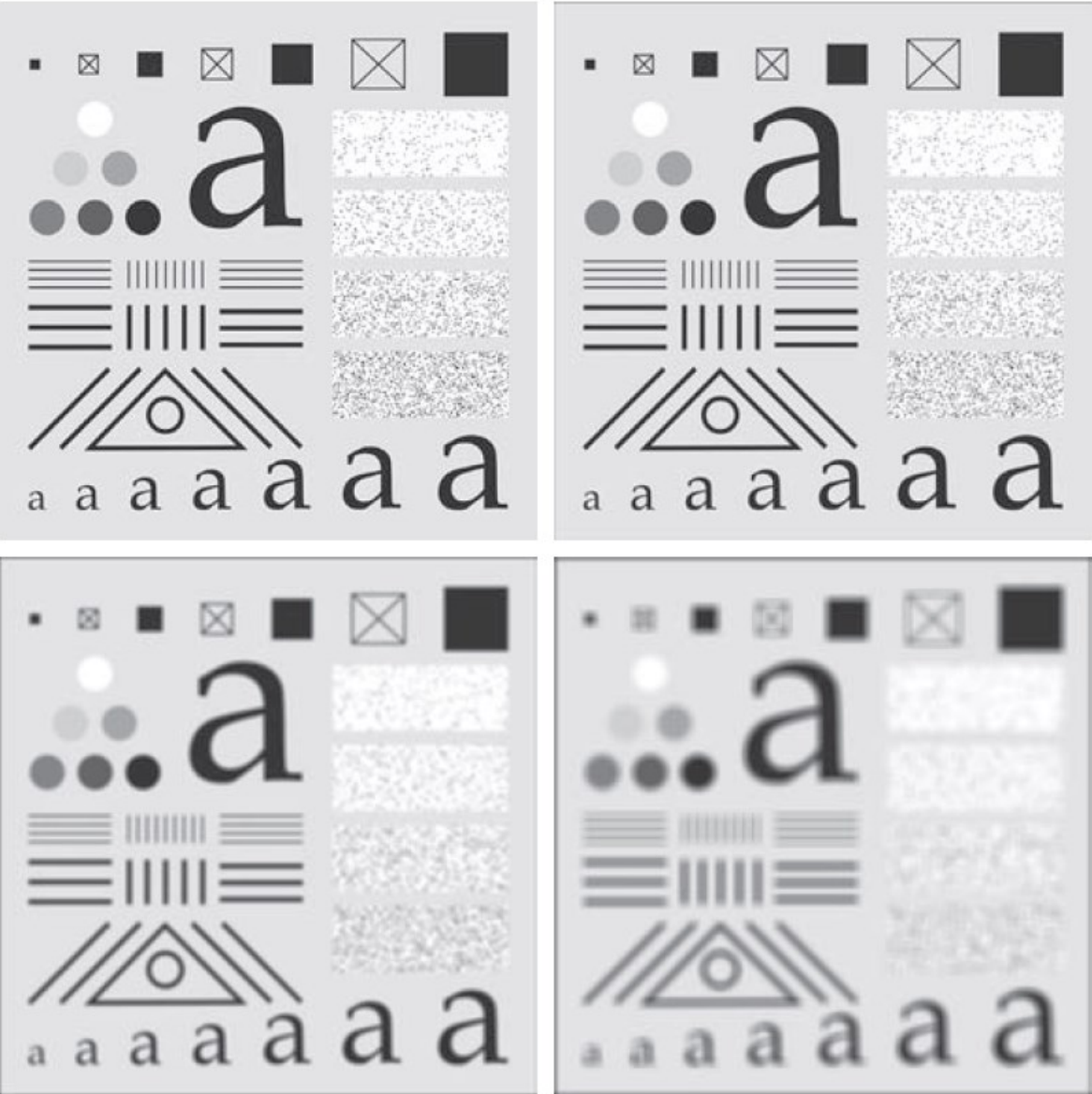
Property	Convolution	Correlation
Commutative	$f \star g = g \star f$	—
Associative	$f \star (g \star h) = (f \star g) \star h$	—
Distributive	$f \star (g + h) = (f \star g) + (f \star h)$	$f \star (g + h) = (f \star g) + (f \star h)$

# Image Processing: Smoothing (Lowpass) Spatial Filters

- Smoothing (averaging) spatial filters are used to **reduce sharp transitions in intensity**.
- An obvious application of smoothing is **noise reduction**.
- Smoothing is used **to reduce irrelevant detail** in an image.

$$\frac{1}{9} \times \begin{array}{|c|c|c|} \hline 1 & 1 & 1 \\ \hline 1 & 1 & 1 \\ \hline 1 & 1 & 1 \\ \hline \end{array}$$

# Image Processing: Smoothing (Lowpass) Spatial Filters



# Image Processing: Sharpening (Highpass) Spatial Filters

- The simplest derivative operator (kernel) is the **Laplacian**, which, for a function (image)  $f(x, y)$  of two variables, is defined as

$$\nabla^2 f = \frac{\partial^2 f}{\partial x^2} + \frac{\partial^2 f}{\partial y^2}$$

- In the  $x$ -direction, we have

$$\frac{\partial^2 f}{\partial x^2} = f(x + 1, y) + f(x - 1, y) - 2f(x, y)$$

- In the  $y$ -direction, we have

$$\frac{\partial^2 f}{\partial y^2} = f(x, y + 1) + f(x, y - 1) - 2f(x, y)$$

# Image Processing: Sharpening (Highpass) Spatial Filters

- It follows from the preceding three equations that the **discrete Laplacian** of two variables is

$$\nabla^2 f(x, y) = f(x + 1, y) + f(x - 1, y) + f(x, y + 1) + f(x, y - 1) - 4f(x, y)$$

0	1	0
1	-4	1
0	1	0

# Image Processing: Sharpening (Highpass) Spatial Filters





# CVIPtools

CVIPtools for Windows - Southern Illinois University Edwardsville

File View Analysis Enhancement Restoration Compression Utilities Arith/Logic Compare Convert Create Enhance Filter Size Stats Help

Column: Row: R: G: B:

CVIPtools Website  
CVIP-ATAT CVIP-FEPC

Delete All Delete  
Lock Input

Stripey.jpg  
Stripey.jpg\_mean23

Stripey.jpg

Enhancement

Histogram/Contrast Pseudocolor Sharpening Smoothing

Spatial Domain Smoothing Parameters

Mean Filter  
 Gaussian Filter  
 Midpoint Filter  
 Contra-harmonic Filter

Mask Size: 7

Stripey.jpg\_mean23

Apply

The screenshot displays the CVIPtools application interface. On the left, a file list shows 'Stripey.jpg' and 'Stripey.jpg\_mean23'. The main workspace is split into two windows: 'Stripey.jpg' (top-left) and 'Stripey.jpg\_mean23' (bottom-right). The 'Stripey.jpg' window shows a clear image of an orange tabby cat sitting on a white plate on a light blue countertop. The 'Stripey.jpg\_mean23' window shows the same image but with a soft, blurred appearance, demonstrating the effect of a mean filter. An 'Enhancement' dialog box is open over the bottom-right window, with the 'Smoothing' tab selected. Under 'Spatial Domain Smoothing', the 'Mean Filter' is selected. The 'Mask Size' is set to 7. The 'Apply' button is visible at the bottom right of the dialog.

# CVIPtools

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File View Analysis Enhancement Restoration Compression Utilities Arith/Logic Compare Convert Create Enhance Filter Size Stats Help

Column: Row: R: G: B:

CVIPtools Website  
CVIP-ATAT CVIP-FEPC

Delete All Delete  
Lock Input

cam.bmp  
cam.bmp\_Lapl3x3\_...

cam.bmp

Laplacian Filter Type 1

	1	2	3
1	0.000	-1.000	0.000
2	-1.000	4.000	-1.000
3	0.000	-1.000	0.000

Keep original image  
Cancel Reset Apply

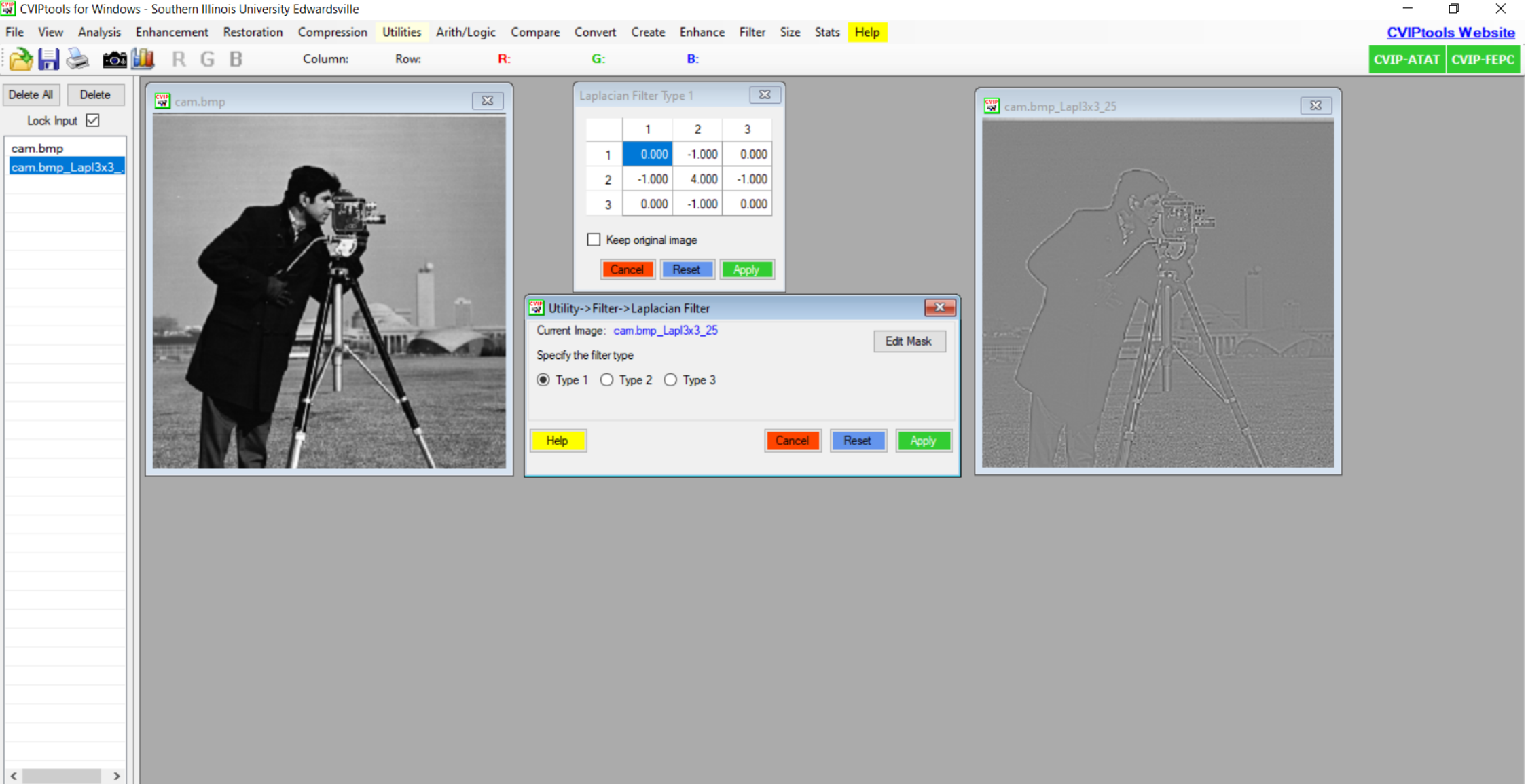
Utility->Filter->Laplacian Filter

Current Image: cam.bmp\_Lapl3x3\_25 Edit Mask

Specify the filter type  
 Type 1  Type 2  Type 3

Help Cancel Reset Apply

cam.bmp\_Lapl3x3\_25



# CVIPtools

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File View Analysis Enhancement Restoration Compression Utilities Arith/Logic Compare Convert Create Enhance Filter Size Stats Help

Column: Row: R: G: B:

CVIPtools Website  
CVIP-ATAT CVIP-FEPC

Delete All Delete  
Lock Input

- butterfly.tif
- butterfly.tif\_Lum34
- butterfly.tif\_Hist36
- butterfly.tif\_Lum34\_H
- butterfly.tif\_Lum34\_T

butterfly.tif

butterfly.tif\_Lum34

butterfly.tif\_Lum34\_Thresh48

butterfly.tif\_Hist36

butterfly.tif\_Lum34\_Hist37

# CVIPtools

CVIPtools for Windows - Southern Illinois University Edwardsville

File View Analysis Enhancement Restoration Compression Utilities Arith/Logic Compare Convert Create Enhance Filter Size Stats Help

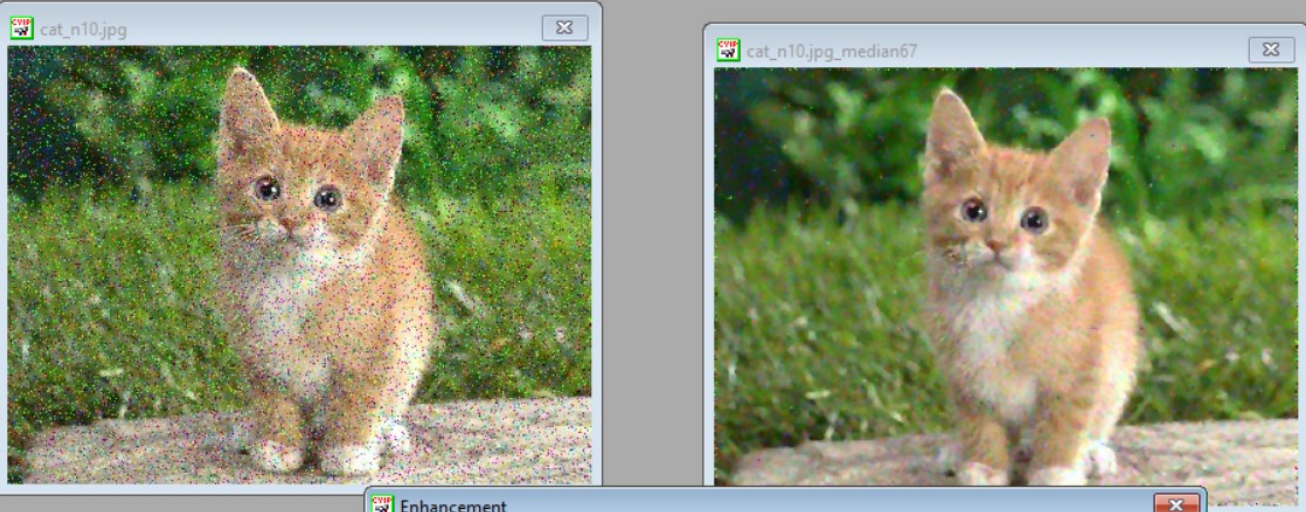
CVIPtools Website

CVIP-ATAT CVIP-FEPC

Delete All Delete

Lock Input

- cat\_n10.jpg
- cat\_n10.jpg\_median



Enhancement

Histogram/Contrast Pseudocolor Sharpening Smoothing

Spatial Domain Smoothing

- Mean Filter
- Gaussian Filter
- Midpoint Filter
- Contra-harmonic Filter
- Yp Mean Filter
- Median Filter
- Kuwahara Filter
- Anisotropic Diffusion Filter

Parameters

Type:  Standard  Adaptive

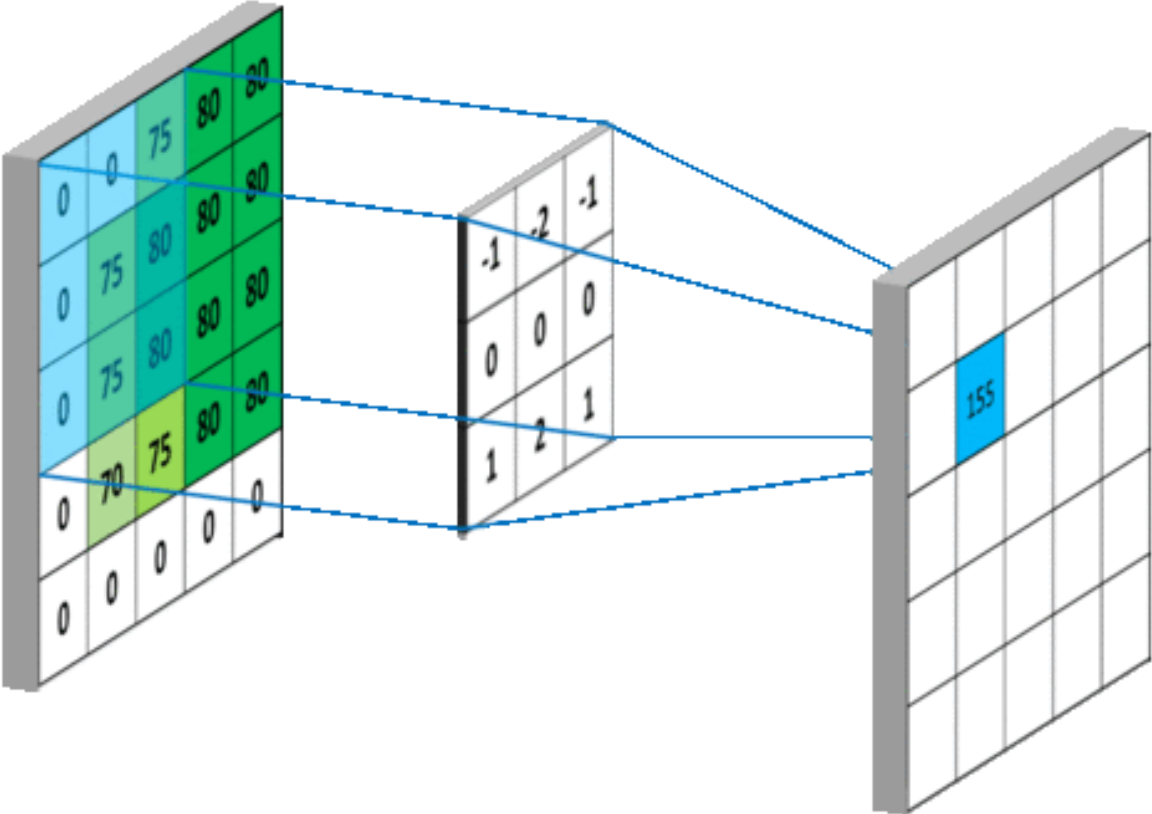
Mask Size: 5

Frequency Domain Smoothing

- FFT Smoothing
- DCT Smoothing

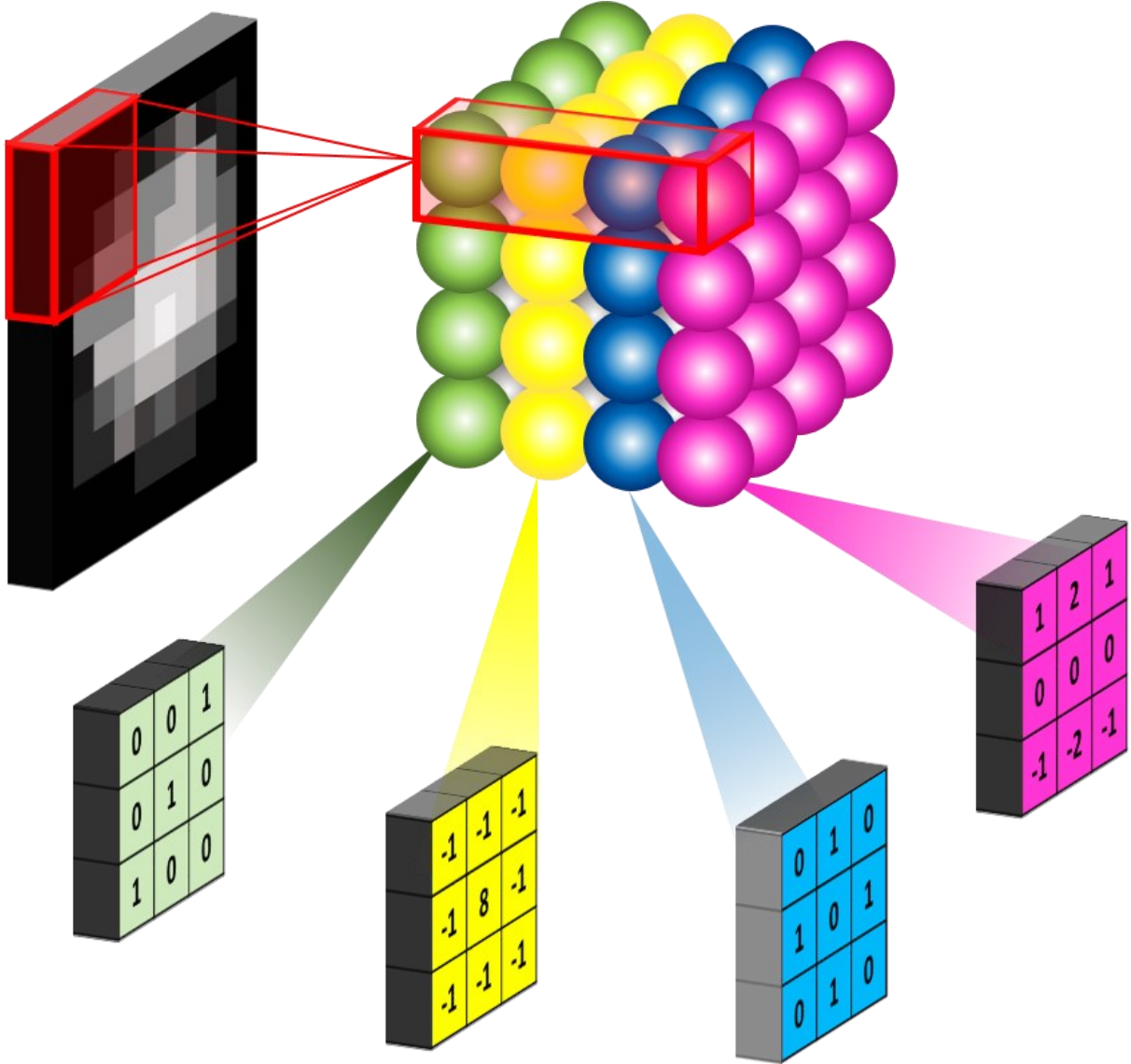
Help Cancel Reset Apply

# Convolutional Neural Networks

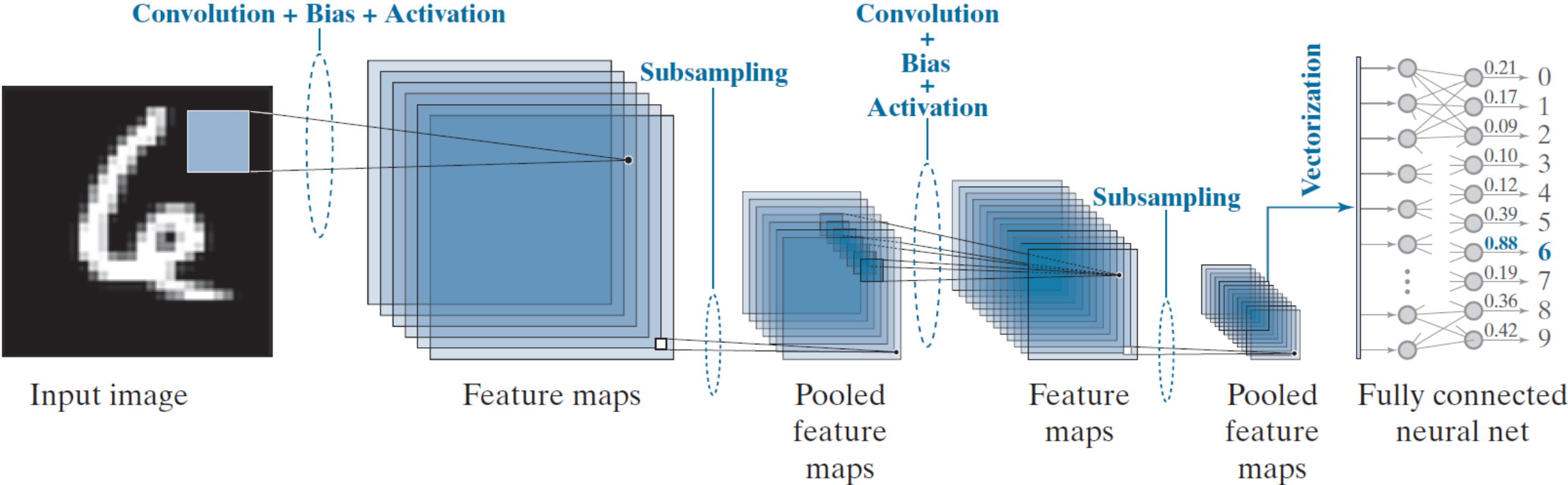


<https://mlnotebook.github.io/post/CNN1/>

# Convolutional Neural Networks



# Convolutional Neural Networks



# Continuous-Time Convolution

- A **convolution** is an **integral** that expresses the amount of overlap of one function when it is **shifted** over another function.

$$y(t) = x(t) * h(t) = \int_{-\infty}^{\infty} x(\lambda) h(t - \lambda) d\lambda$$

$$y(t) = h(t) * x(t) = \int_{-\infty}^{\infty} h(\lambda) x(t - \lambda) d\lambda$$



# Discrete-Time Convolution

- The output signal  $y[n]$  of a **DTLTI system** is obtained by **convolving** the **input signal**  $x[n]$  and the **impulse response**  $h[n]$  of the system.
- This relationship is expressed in compact notation as

$$y[n] = x[n] * h[n]$$

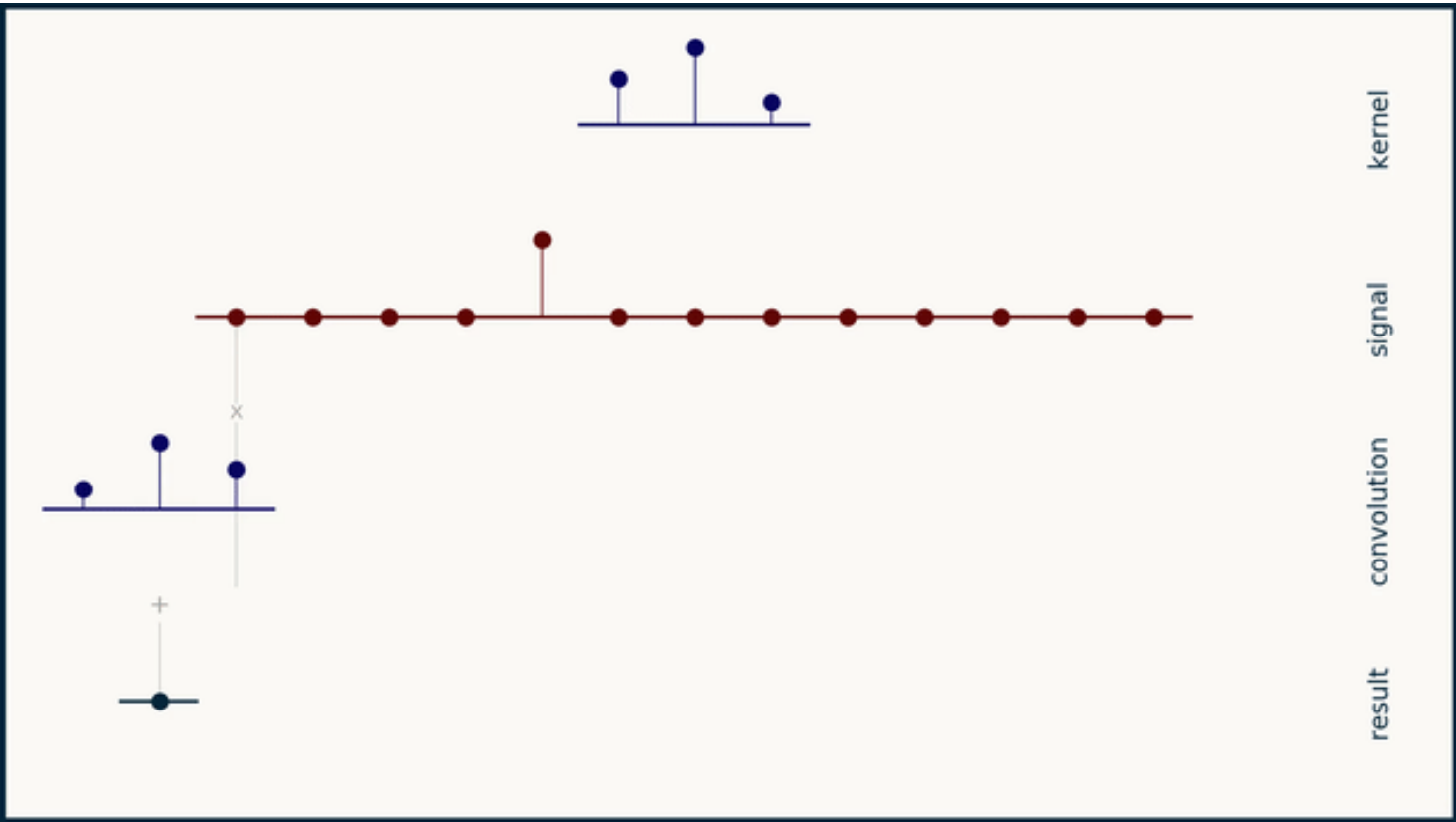
where the symbol  $*$  represents the **convolution operator**.

$$y[n] = x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k] h[n - k]$$

$$y[n] = h[n] * x[n] = \sum_{k=-\infty}^{\infty} h[k] x[n - k]$$

# Discrete-Time Convolution

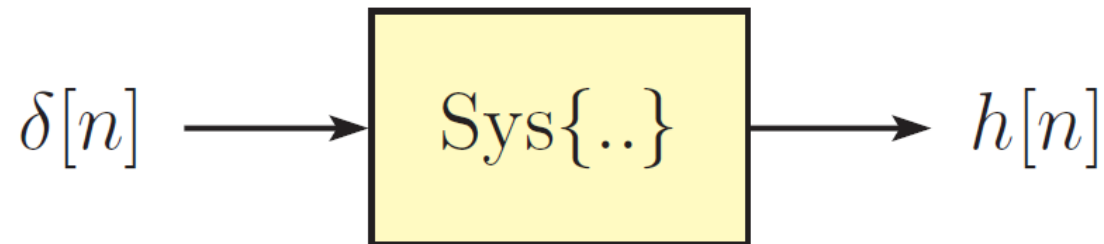
$$y[n] = x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k] h[n - k]$$



[https://e2eml.school/convolution\\_one\\_d.html](https://e2eml.school/convolution_one_d.html)

# Impulse Response

- A constant-coefficient linear **difference equation** is sufficient for describing a DTLTI system.
- The **impulse response** also constitutes a **complete description** of a **DTLTI system**.
- The response of a DTLTI system to any arbitrary input signal  $x[n]$  can be uniquely determined from the **knowledge of its impulse response**.



## Example 3.19

**Example 3.19:** A simple discrete-time convolution example

A discrete-time system is described through the impulse response

$$h[n] = \left\{ \underset{\substack{\uparrow \\ n=0}}{4}, 3, 2, 1 \right\}$$

Use the convolution operation to find the response of the system to the input signal

$$x[n] = \left\{ \underset{\substack{\uparrow \\ n=0}}{-3}, 7, 4 \right\}$$

## Example 3.19 – Solution

**Solution:** Consider the convolution sum given by Eqn. (3.128). Let us express the terms inside the convolution summation, namely  $x[k]$  and  $h[n - k]$ , as functions of  $k$ .

$$x[k] = \{ \underset{\substack{\uparrow \\ k=0}}{-3}, 7, 4 \}$$

$$h[-k] = \{ 1, 2, 3, \underset{\substack{\uparrow \\ k=0}}{4} \}$$

$$h[n - k] = \{ 1, 2, 3, \underset{\substack{\uparrow \\ k=n}}{4} \}$$

In its general form both limits of the summation in Eqn. (3.128) are infinite. On the other hand,  $x[k] = 0$  for negative values of the summation index  $k$ , so setting the lower limit of the summation to  $k = 0$  would have no effect on the result. Similarly, the last significant sample of  $x[k]$  is at index  $k = 2$ , so the upper limit can be changed to  $k = 2$  without affecting the result as well, leading to

$$y[n] = \sum_{k=0}^2 x[k] h[n - k] \quad (3.131)$$

## Example 3.19 – Solution

For  $n = 0$ :

$$\begin{aligned}y[0] &= \sum_{k=0}^0 x[k] h[0 - k] \\ &= x[0] h[0] = (-3) (4) = -12\end{aligned}$$

For  $n = 1$ :

$$\begin{aligned}y[1] &= \sum_{k=0}^1 x[k] h[1 - k] \\ &= x[0] h[1] + x[1] h[0] \\ &= (-3) (3) + (7) (4) = 19\end{aligned}$$

## Example 3.19 – Solution

For  $n = 2$ :

$$\begin{aligned}y[2] &= \sum_{k=0}^2 x[k] h[2-k] \\ &= x[0] h[2] + x[1] h[1] + x[2] h[0] \\ &= (-3) (2) + (7) (3) + (4) (4) = 31\end{aligned}$$

For  $n = 3$ :

$$\begin{aligned}y[3] &= \sum_{k=0}^2 x[k] h[3-k] \\ &= x[0] h[3] + x[1] h[2] + x[2] h[1] \\ &= (-3) (1) + (7) (2) + (4) (3) = 23\end{aligned}$$

## Example 3.19 – Solution

For  $n = 4$ :

$$\begin{aligned}y[4] &= \sum_{k=1}^2 x[k] h[4 - k] \\ &= x[1] h[3] + x[2] h[2] \\ &= (7) (1) + (4) (2) = 15\end{aligned}$$

For  $n = 5$ :

$$\begin{aligned}y[5] &= \sum_{k=2}^2 x[k] h[5 - k] \\ &= x[2] h[3] = (4) (1) = 4\end{aligned}$$

Thus the convolution result is

$$y[n] = \{ -12, 19, 31, 23, 15, 4 \}$$

$\uparrow$   
 $n=0$



# Convolution Using MATLAB

```
>> x = [-3, 7, 4];
```

```
>> h = [4, 3, 2, 1];
```

```
>> y = conv(x, h)
```

```
y =
```

```
    -12    19    31    23    15     4
```

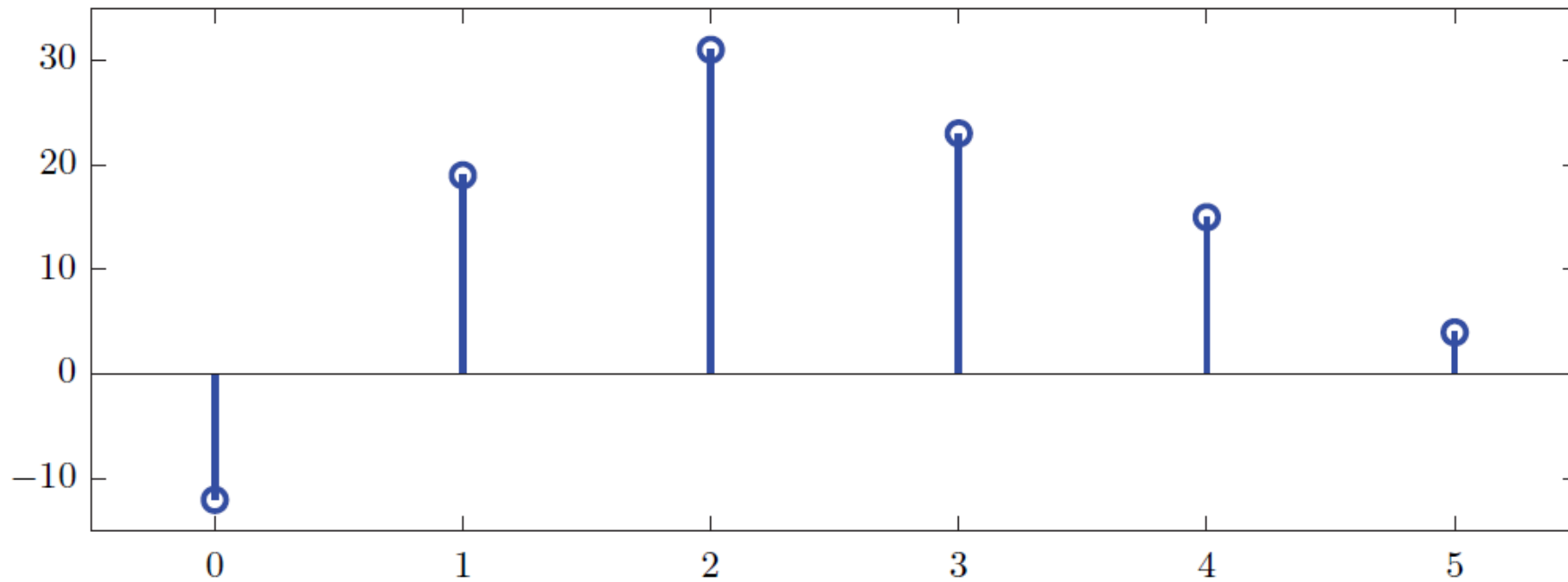
```
>> y = conv(h, x)
```

```
y =
```

```
    -12    19    31    23    15     4
```

# Convolution Using MATLAB

```
>> n = [0:5];  
>> stem(n, y);
```



## Problem 3.5

- 3.5.** The response of a linear and time-invariant system to the input signal  $x[n] = \delta[n]$  is given by

$$\text{Sys} \{ \delta[n] \} = \left\{ \underset{\substack{\uparrow \\ n=0}}{2}, 1, -1 \right\}$$

Determine the response of the system to the following input signals:

- a.**  $x[n] = \delta[n] + \delta[n - 1]$
- b.**  $x[n] = \delta[n] - 2\delta[n - 1] + \delta[n - 2]$
- c.**  $x[n] = u[n] - u[n - 5]$

# Problem 3.5 (a) – Solution

$$x[n] = \delta[n] + \delta[n - 1]$$

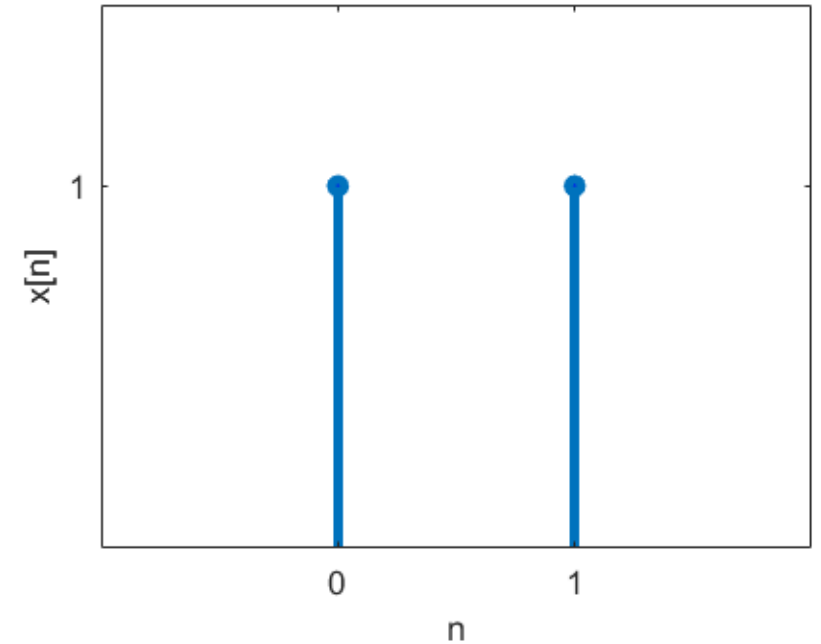
$$x[n] = \left\{ \underset{\substack{\uparrow \\ n=0}}{1}, 1 \right\}$$

$$h[n] = \left\{ \underset{\substack{\uparrow \\ n=0}}{2}, 1, -1 \right\}$$

$$x[k] = \left\{ \underset{\substack{\uparrow \\ k=0}}{1}, 1 \right\}$$

$$h[n - k] = \left\{ -1, 1, \underset{\substack{\uparrow \\ k=n}}{2} \right\}$$

$$y[n] = x[n] * h[n] = \left\{ \underset{\substack{\uparrow \\ n=0}}{2}, 3, 0, -1 \right\}$$



## Problem 3.5 (a) – Another Solution

**a.** Since the system is linear

$$y[n] = \text{Sys} \{ \delta[n] + \delta[n-1] \} = \text{Sys} \{ \delta[n] \} + \text{Sys} \{ \delta[n-1] \}$$

The system is also time-invariant, therefore

$$\text{Sys} \{ \delta[n] \} = \{ \underset{\substack{\uparrow \\ n=0}}{2}, 1, -1 \} \quad \Rightarrow \quad \text{Sys} \{ \delta[n-1] \} = \{ \underset{\substack{\uparrow \\ n=0}}{0}, 2, 1, -1 \}$$

and

$$y[n] = \{ \underset{\substack{\uparrow \\ n=0}}{2}, 3, 0, -1 \}$$

# Problem 3.5 (b) – Solution

$$x[n] = \delta[n] - 2\delta[n-1] + \delta[n-2]$$

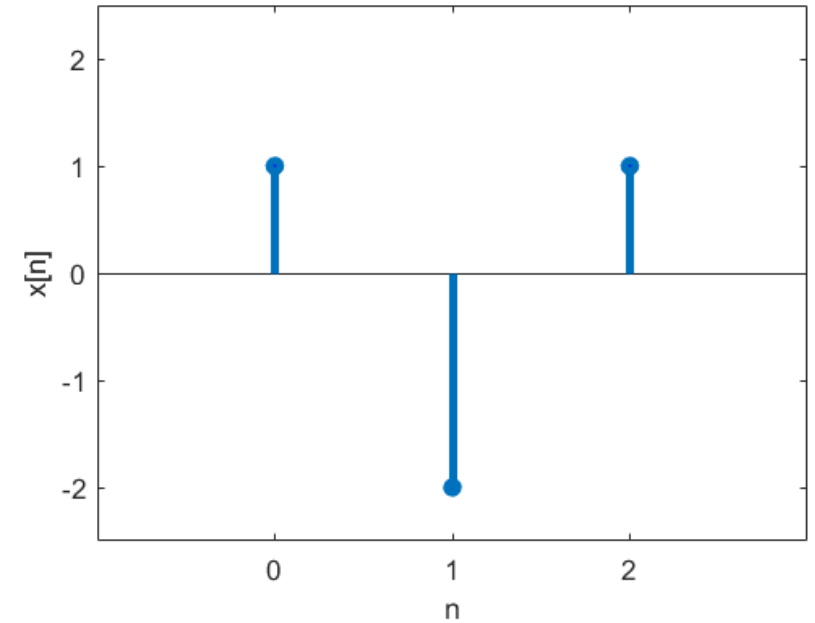
$$x[n] = \left\{ \underset{\substack{\uparrow \\ n=0}}{1}, -2, 1 \right\}$$

$$h[n] = \left\{ \underset{\substack{\uparrow \\ n=0}}{2}, 1, -1 \right\}$$

$$x[k] = \left\{ \underset{\substack{\uparrow \\ k=0}}{1}, -2, 1 \right\}$$

$$h[n-k] = \left\{ -1, 1, \underset{\substack{\uparrow \\ k=n}}{2} \right\}$$

$$y[n] = x[n] * h[n] = \left\{ \underset{\substack{\uparrow \\ n=0}}{2}, -3, -1, 3, -1 \right\}$$



## Problem 3.5 (b) – Another Solution

**b.** Since the system is linear

$$\begin{aligned}y[n] &= \text{Sys} \{ \delta[n] - 2\delta[n-1] + \delta[n-2] \} \\ &= \text{Sys} \{ \delta[n] \} - 2 \text{Sys} \{ \delta[n-1] \} + \text{Sys} \{ \delta[n-2] \}\end{aligned}$$

The system is also time-invariant, therefore

$$\text{Sys} \{ \delta[n-1] \} = \{ \underset{\substack{\uparrow \\ n=0}}{0}, 2, 1, -1 \}$$

and

$$\text{Sys} \{ \delta[n-2] \} = \{ \underset{\substack{\uparrow \\ n=0}}{0}, 0, 2, 1, -1 \}$$

The response is

$$y[n] = \{ \underset{\substack{\uparrow \\ n=0}}{2}, -3, -1, 3, -1 \}$$

# Problem 3.5 (c) – Solution

$$x[n] = u[n] - u[n - 5]$$

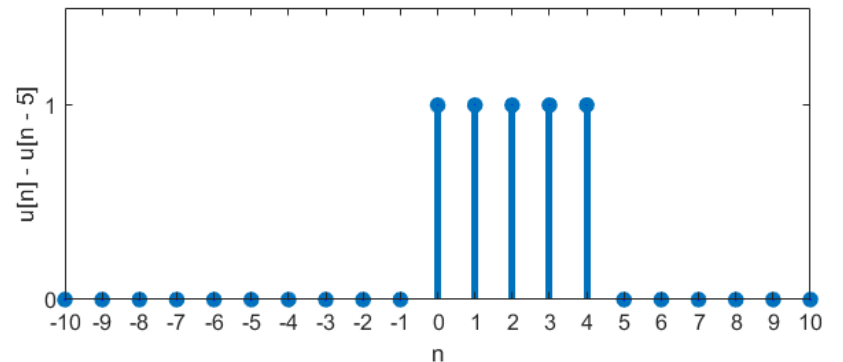
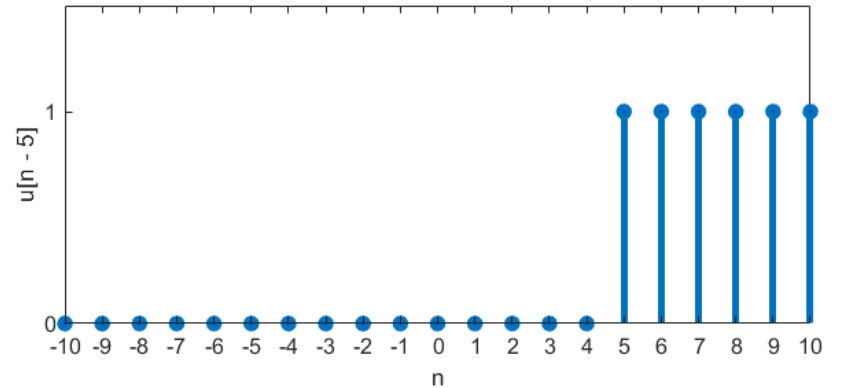
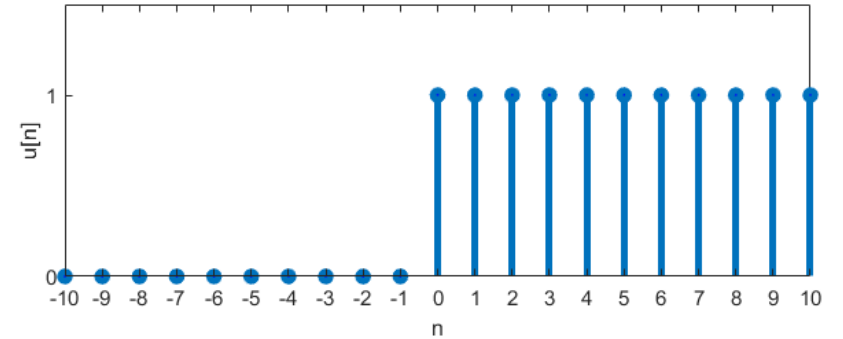
$$x[n] = \{ \underset{\substack{\uparrow \\ n=0}}{1}, 1, 1, 1, 1 \}$$

$$h[n] = \{ \underset{\substack{\uparrow \\ n=0}}{2}, 1, -1 \}$$

$$x[k] = \{ \underset{\substack{\uparrow \\ k=0}}{1}, 1, 1, 1, 1 \}$$

$$h[n - k] = \{ -1, 1, \underset{\substack{\uparrow \\ k=n}}{2} \}$$

$$y[n] = x[n] * h[n] = \{ \underset{\substack{\uparrow \\ n=0}}{2}, 3, 2, 2, 2, 0, -1 \}$$





# Problem 3.5 (c) – Another Solution

c.

$$u[n] - u[n-5] = \delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3] + \delta[n-4]$$

Using the linearity of the system we have

$$\begin{aligned} y[n] &= \text{Sys} \{ \delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3] + \delta[n-4] \} \\ &= \text{Sys} \{ \delta[n] \} + \text{Sys} \{ \delta[n-1] \} + \text{Sys} \{ \delta[n-2] \} + \text{Sys} \{ \delta[n-3] \} + \text{Sys} \{ \delta[n-4] \} \end{aligned}$$

Since the system is also time-invariant, we have

$$\text{Sys} \{ \delta[n-1] \} = \{ \underset{\substack{\uparrow \\ n=0}}{0}, 2, 1, -1 \}$$

$$\text{Sys} \{ \delta[n-2] \} = \{ \underset{\substack{\uparrow \\ n=0}}{0}, 0, 2, 1, -1 \}$$

$$\text{Sys} \{ \delta[n-3] \} = \{ \underset{\substack{\uparrow \\ n=0}}{0}, 0, 0, 2, 1, -1 \}$$

$$\text{Sys} \{ \delta[n-4] \} = \{ \underset{\substack{\uparrow \\ n=0}}{0}, 0, 0, 0, 2, 1, -1 \}$$

The output signal is

$$y[n] = \{ \underset{\substack{\uparrow \\ n=0}}{2}, 3, 2, 2, 2, 0, -1 \}$$

# Causality in Discrete-Time Systems

- A system is said to be **causal** if the current value of the output signal **depends only on current and past values** of the input signal, **but not on its future values**.

- A discrete-time system defined by the relationship is **causal**

$$y[n] = y[n - 1] + x[n] - 3x[n - 1]$$

- A discrete-time system defined by the relationship is **non-causal**

$$y[n] = y[n - 1] + x[n] - 3x[n + 1]$$

# Stability in Discrete-Time Systems

- A system is said to be **stable** in the **bounded-input bounded-output (BIBO)** sense if any bounded input signal is guaranteed to produce a bounded output signal.
- A discrete-time input signal  $x[n]$  is said to be **bounded** if an **upper bound**  $B_x$  exists such that

$$|x[n]| < B_x < \infty \quad \text{implies that} \quad |y[n]| < B_y < \infty$$

# Causality and Stability in DTLTI Systems

- The impulse response of a **causal** DTLTI should be **equal to zero for all negative index values**.

$$h[n] = 0 \quad \text{for all } n < 0$$

- For a DTLTI system to be **stable**, its **impulse response** must be **absolute summable**.

$$\sum_{k=-\infty}^{\infty} |h[k]| < \infty$$

## Problem 3.25

**3.25.** Determine whether the system is causal and/or stable.

**a.** 
$$y[n] = \text{Sys} \{x[n]\} = \sum_{k=-\infty}^n x[k]$$

**c.** 
$$y[n] = \text{Sys} \{x[n]\} = \sum_{k=0}^n x[k] \quad \text{for } n \geq 0$$

**e.** 
$$y[n] = \text{Sys} \{x[n]\} = \sum_{k=n-10}^{n+10} x[k]$$

## Problem 3.25 (a) – Solution

$$\mathbf{a.} \quad y[n] = \text{Sys} \{x[n]\} = \sum_{k=-\infty}^n x[k]$$

$$\text{Sys} \{\delta[n]\} = \sum_{k=-\infty}^n \delta[k] = \begin{cases} 1, & n \geq 0 \\ 0, & n < 0 \end{cases}$$

Therefore,  $h[n] = u[n]$ .

Since  $h[n] = 0$  for all  $n < 0$ , the system is causal.

$$\sum_{k=-\infty}^{\infty} |h[k]| = \sum_{k=0}^{\infty} 1 \rightarrow \infty$$

So, the system is not stable.

## Problem 3.25 (c) – Solution

$$\mathbf{c.} \quad y[n] = \text{Sys} \{x[n]\} = \sum_{k=0}^n x[k] \quad \text{for } n \geq 0$$

$$\text{Sys} \{\delta[n]\} = \sum_{k=0}^n \delta[k] = 1 \quad \text{for } n \geq 0$$

Therefore,  $h[n] = u[n]$ .

Since  $h[n] = 0$  for all  $n < 0$ , the system is causal.

$$\sum_{k=-\infty}^{\infty} |h[k]| = \sum_{k=0}^{\infty} 1 \rightarrow \infty$$

So, the system is not stable.

## Problem 3.25 (e) – Solution

$$\mathbf{e.} \quad y[n] = \text{Sys} \{x[n]\} = \sum_{k=n-10}^{n+10} x[k]$$

$$\text{Sys} \{\delta[n]\} = \sum_{k=n-10}^{n+10} \delta[k] = \begin{cases} 1, & n-10 \leq 0 \leq n+10 \\ 0, & \text{otherwise} \end{cases} \Rightarrow -10 \leq n \leq 10$$

Therefore  $h[n] = u[n+10] - u[n-11]$ .

Since  $h[n] \neq 0$  for all  $n < 0$ , the system is not causal.

$$\sum_{k=-\infty}^{\infty} |h[k]| = \sum_{k=-10}^{10} 1 < \infty$$

So, the system is stable.